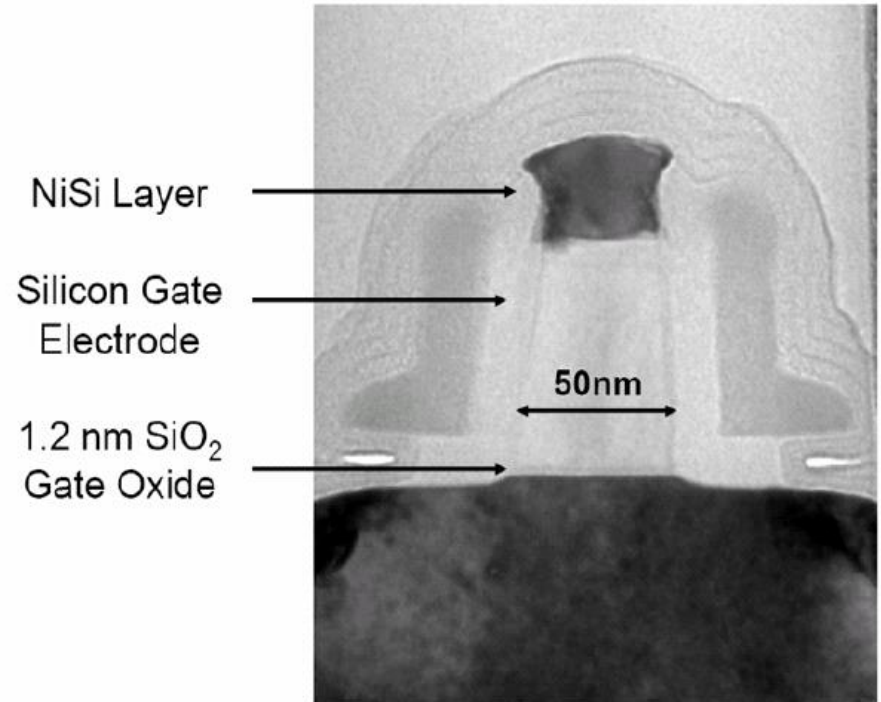
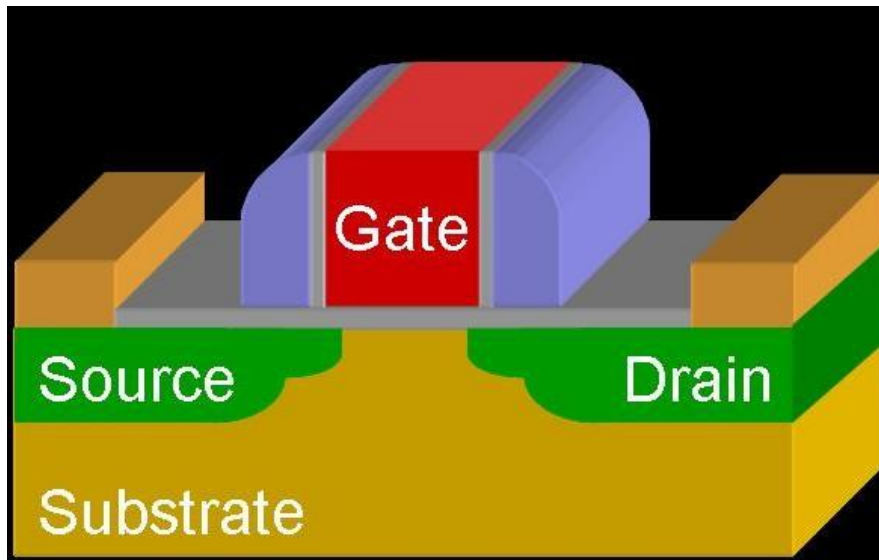
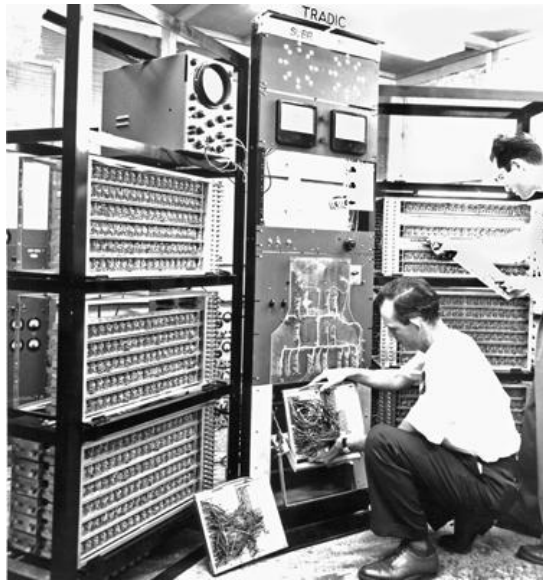
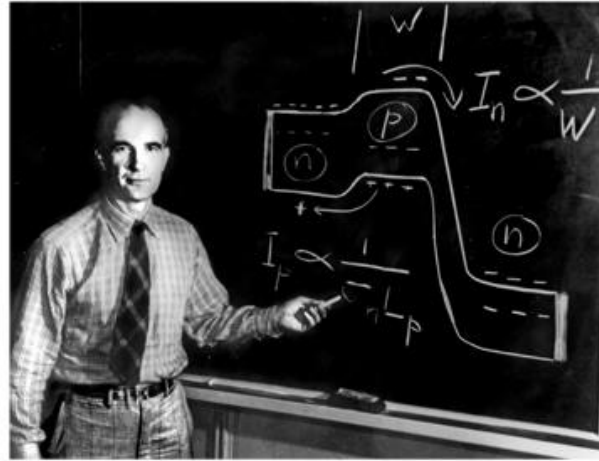
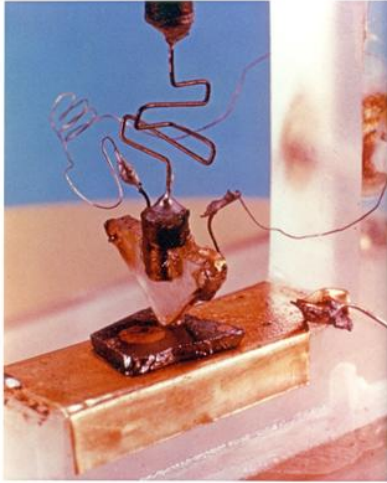


MOSFETs



A little bit of history..



Aug. 27, 1963

DAWON KAHNG

3,102,230

ELECTRIC FIELD CONTROLLED SEMICONDUCTOR DEVICE

Filed May 31, 1960

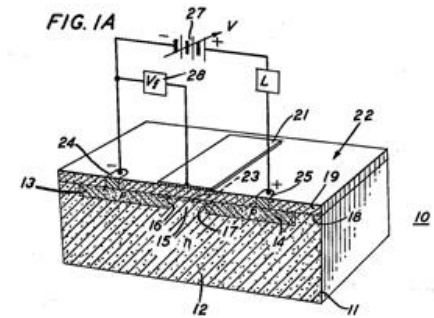
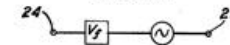


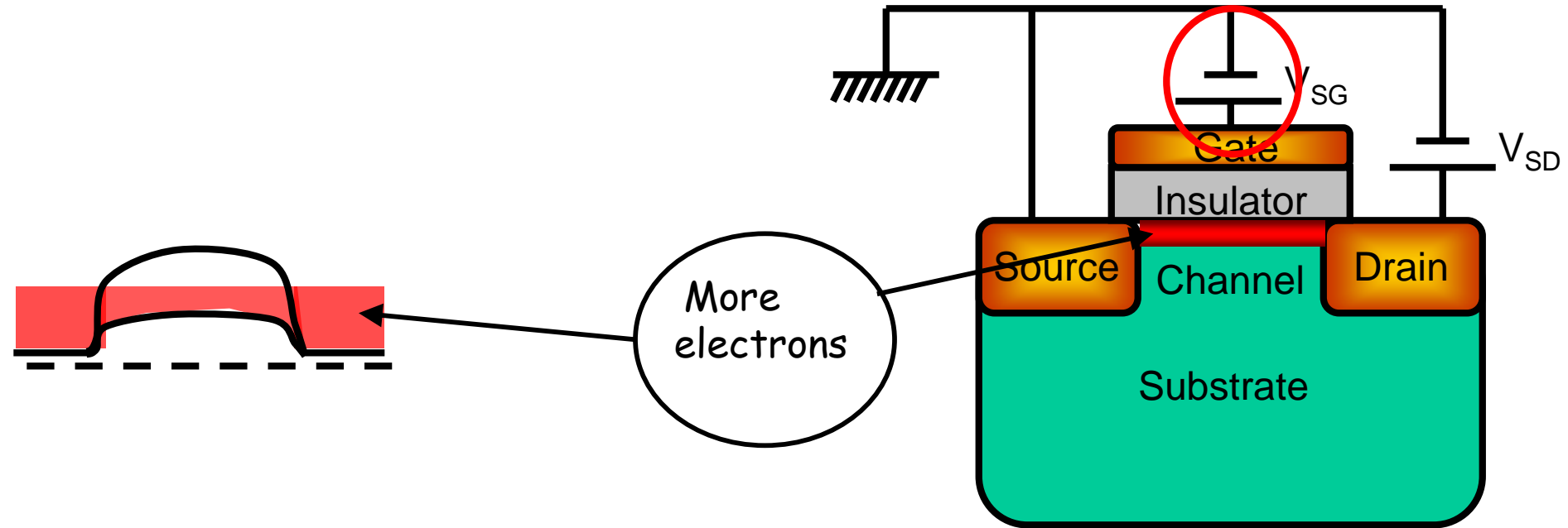
FIG. 1B



Operation of a transistor

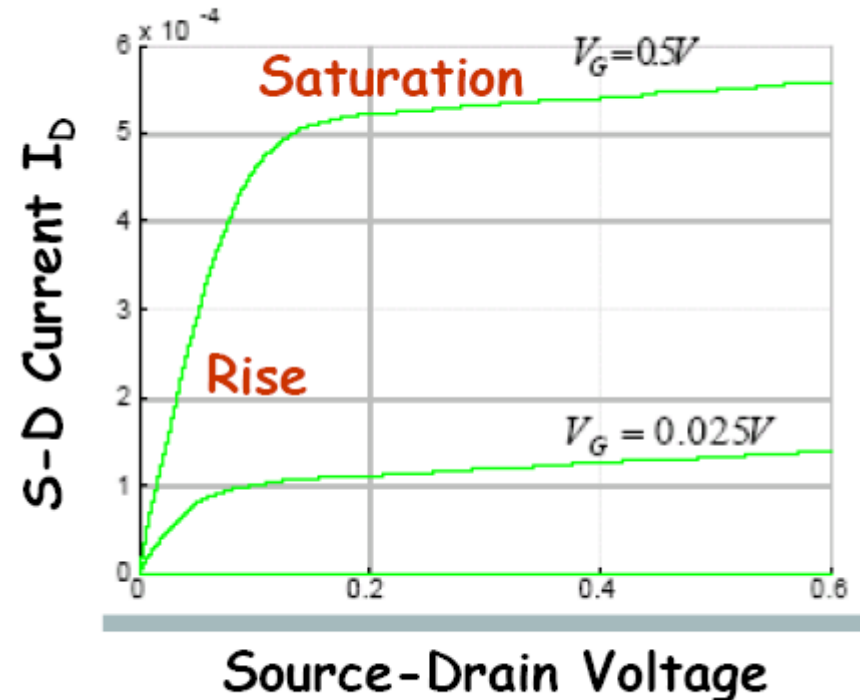
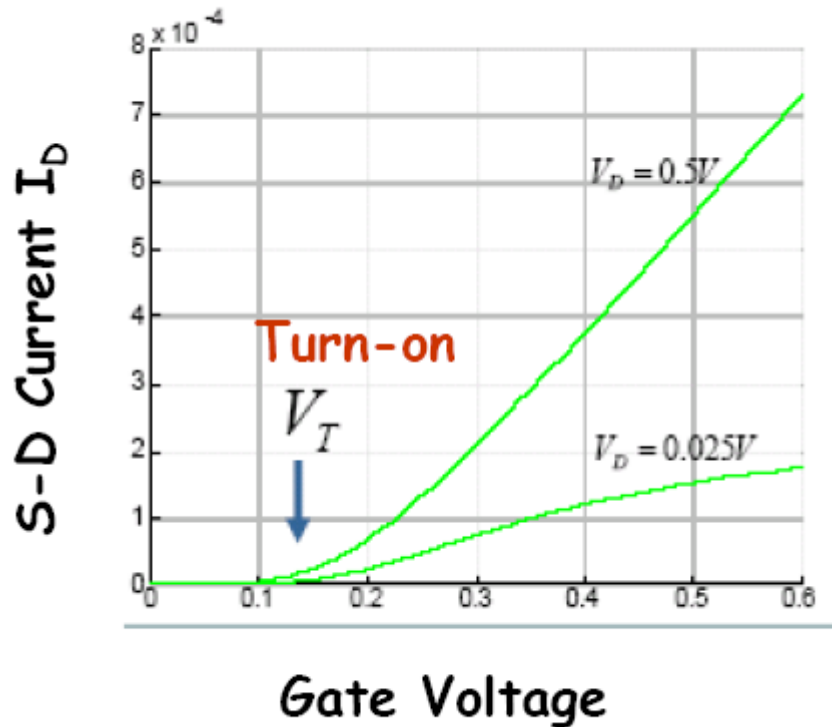
$$V_{SG} > 0$$

n type operation



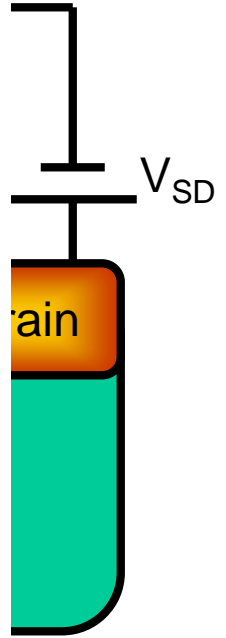
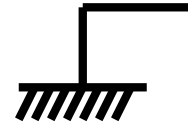
Positive gate bias attracts electrons into channel
Channel now becomes more conductive

Operation of a transistor

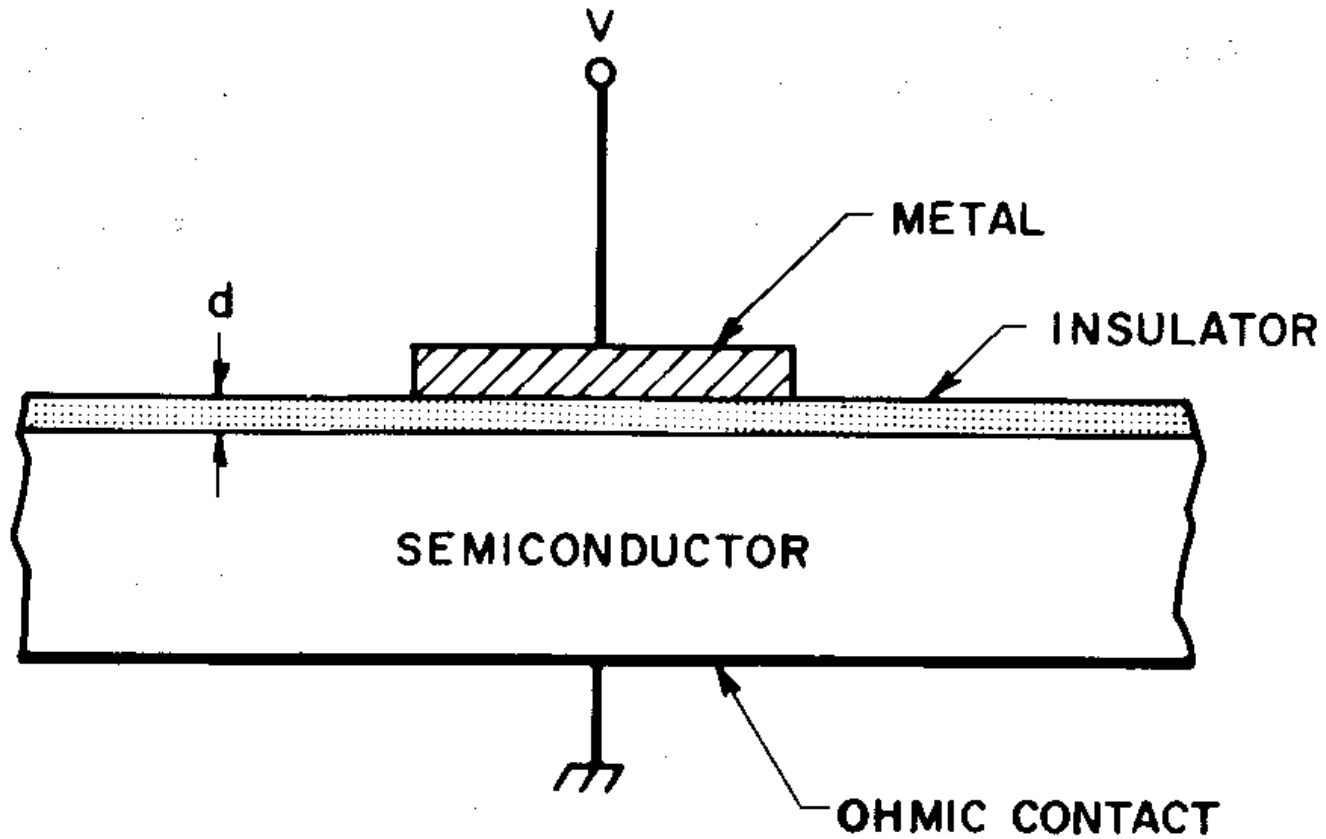


Transistor turns on at high gate voltage
Transistor current saturates at high drain bias

Start with a MOS capacitor



MIS Diode (MOS capacitor) - Ideal



Questions



✓ What is the MOS capacitance? $Q_S(\Psi_S)$

What are the local conditions during inversion? $\Psi_{S,cr}$

How does the potential vary with position? $\Psi(x)$

How much inversion charge is generated at the surface? $Q_{inv}(x, \Psi_S)$

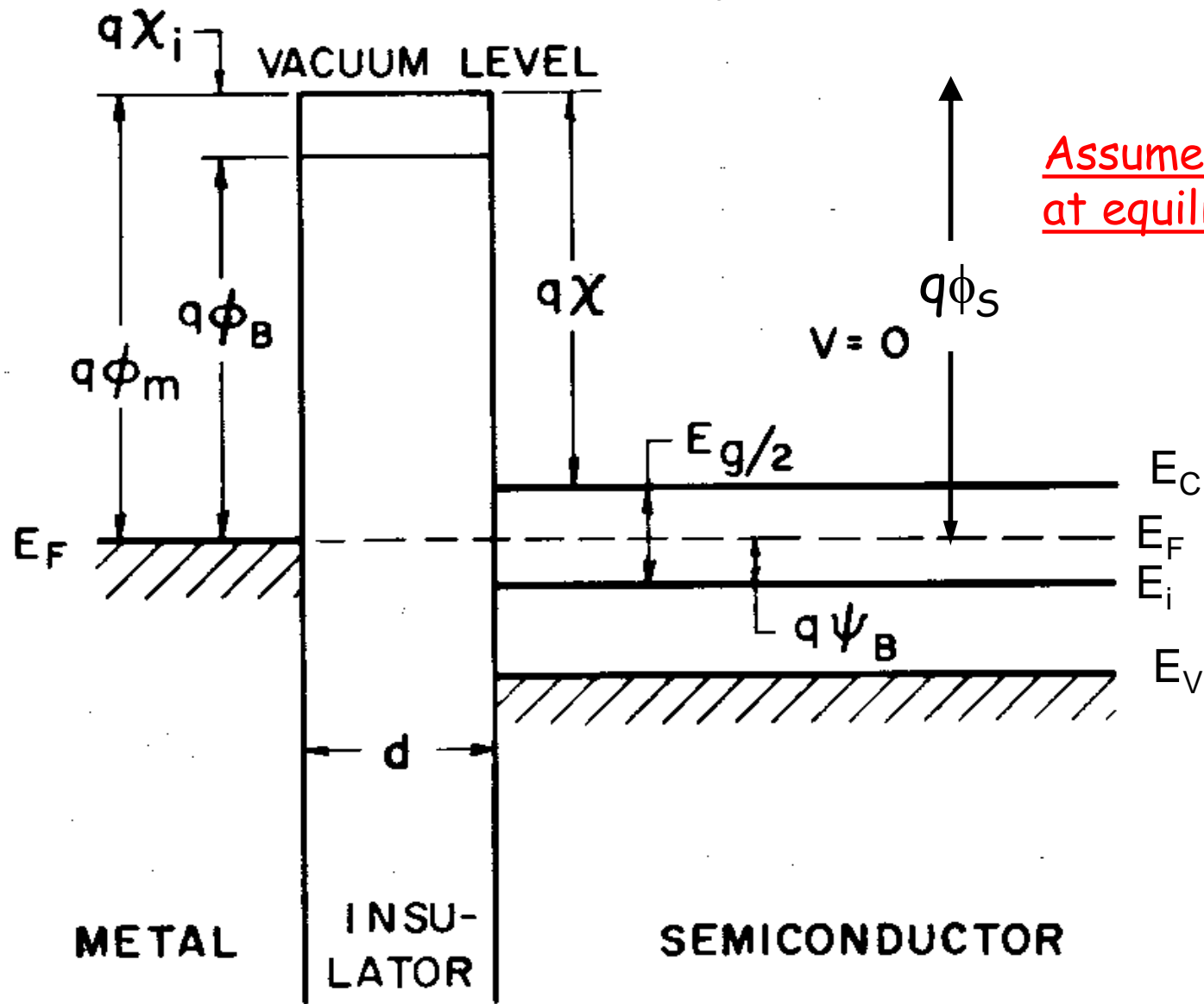
Add in the oxide: how does the voltage divide? $\Psi_S(V_G)$, $\Psi_{ox}(V_G)$

How much gate voltage do you need to invert the channel? V_{TH}

How much inversion charge is generated by the gate? $Q_{inv}(V_G)$

What's the overall C-V of the MOSFET? $Q_S(V_G)$

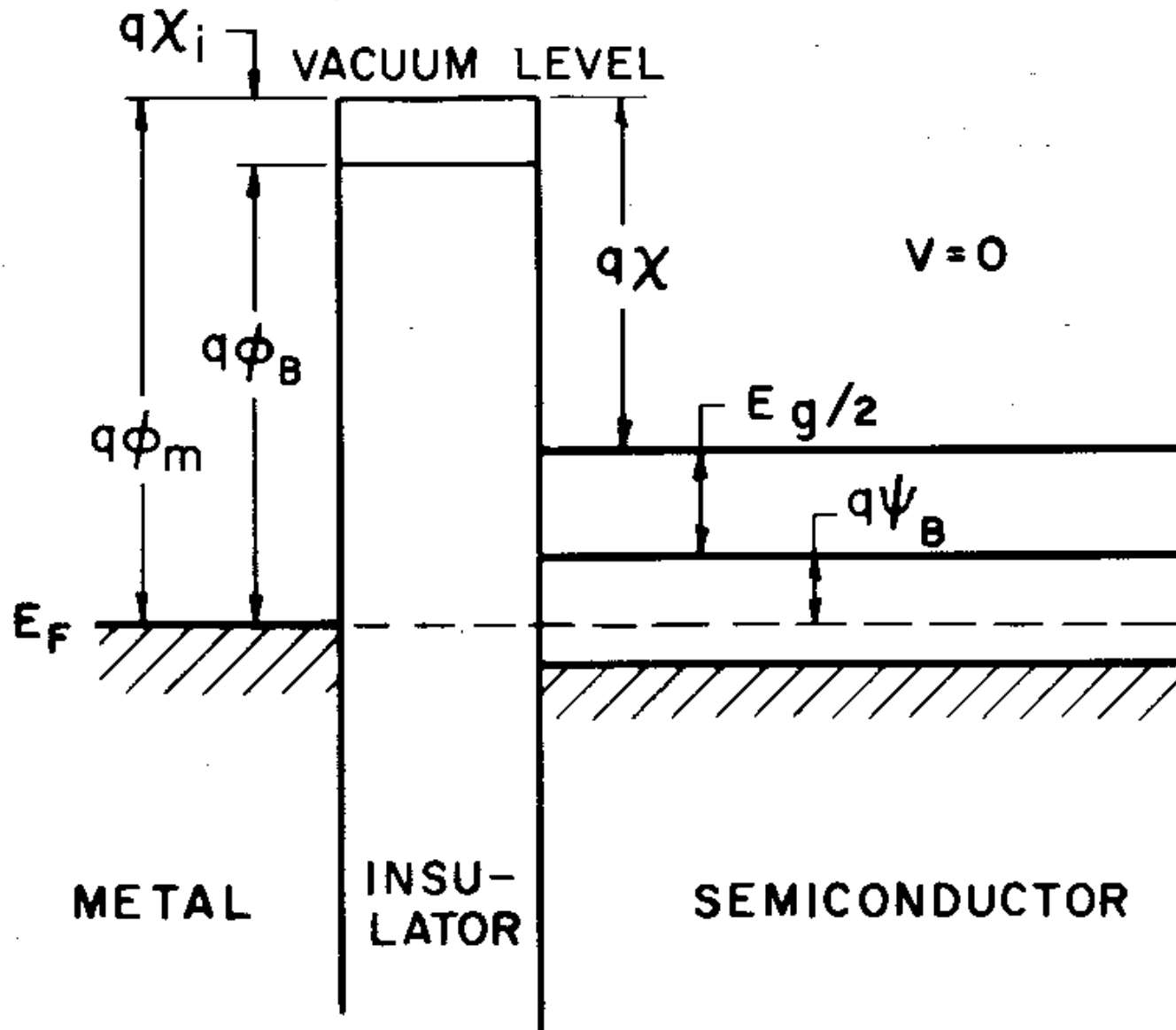
Ideal MIS Diode n-type, $V_{\text{appl}}=0$



Ideal MIS Diode n-type, $V_{\text{appl}}=0$

$$\phi_{ms} \equiv \phi_m - \left(\chi + \frac{E_g}{2q} - \psi_B \right) = 0$$

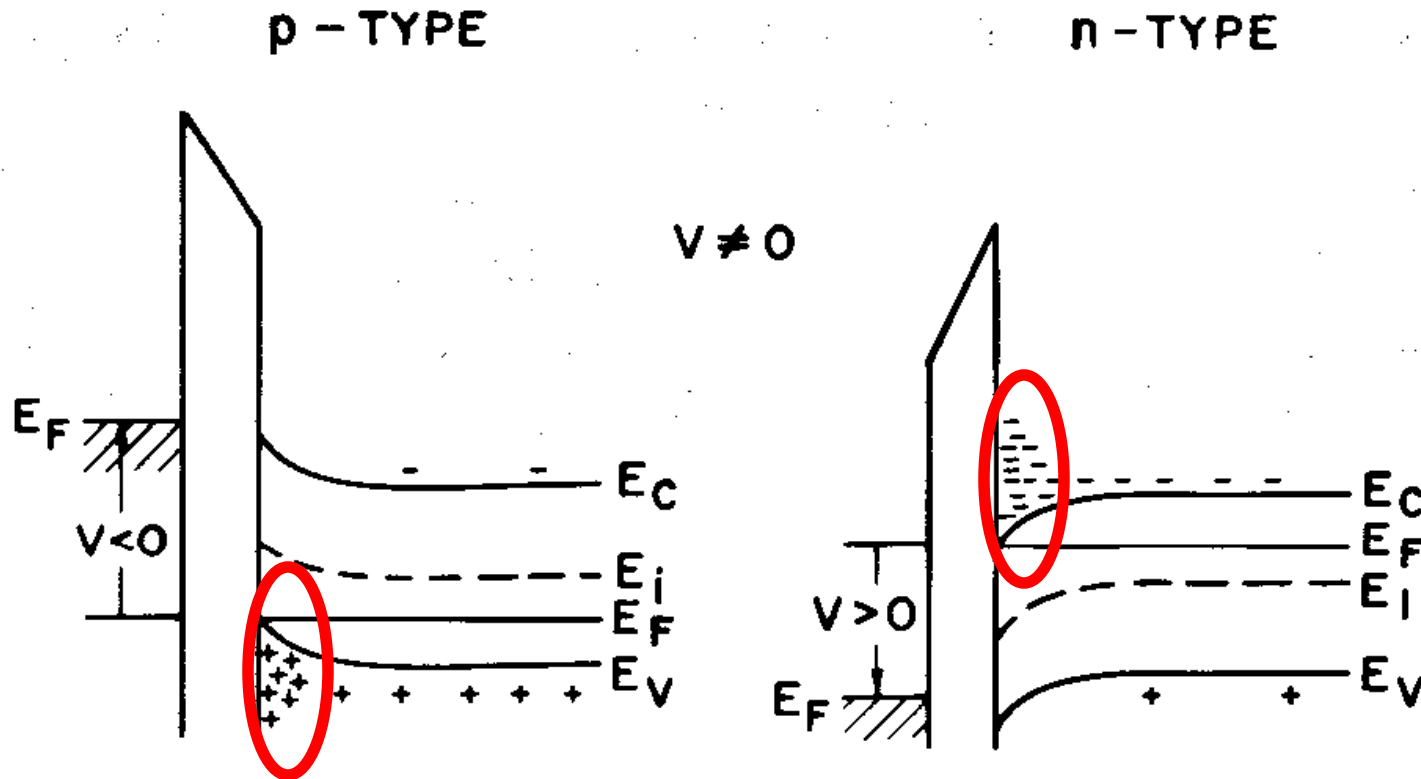
Ideal MIS Diode p-type, $V_{\text{appl}}=0$



Ideal MIS Diode p-type, $V_{\text{appl}}=0$

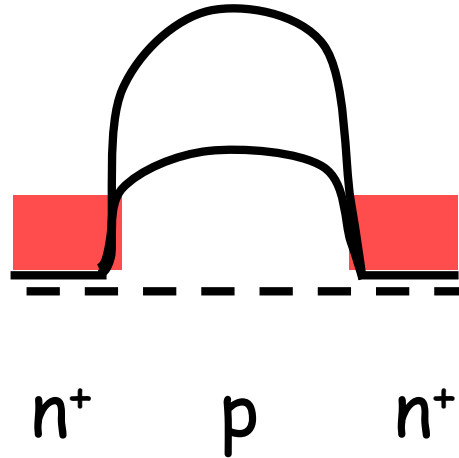
$$\phi_{ms} \equiv \phi_m - \left(\chi + \frac{E_g}{2q} + \psi_B \right) = 0$$

Accumulation



Pulling in majority carriers at surface

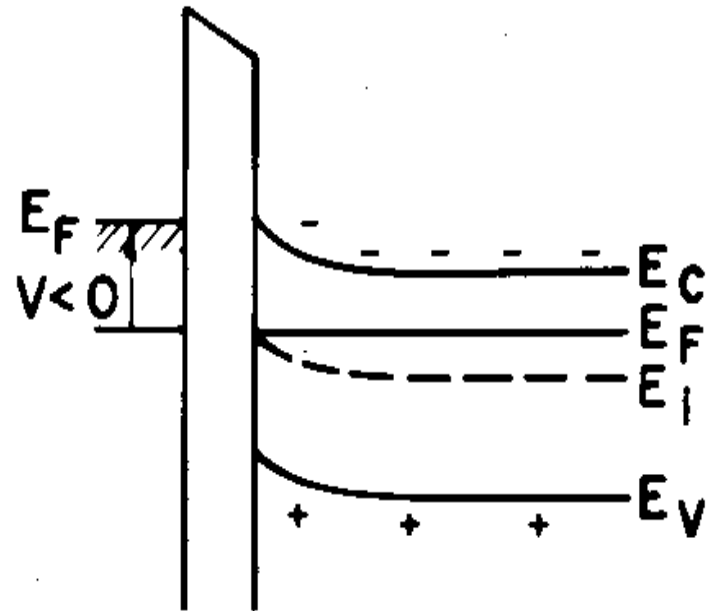
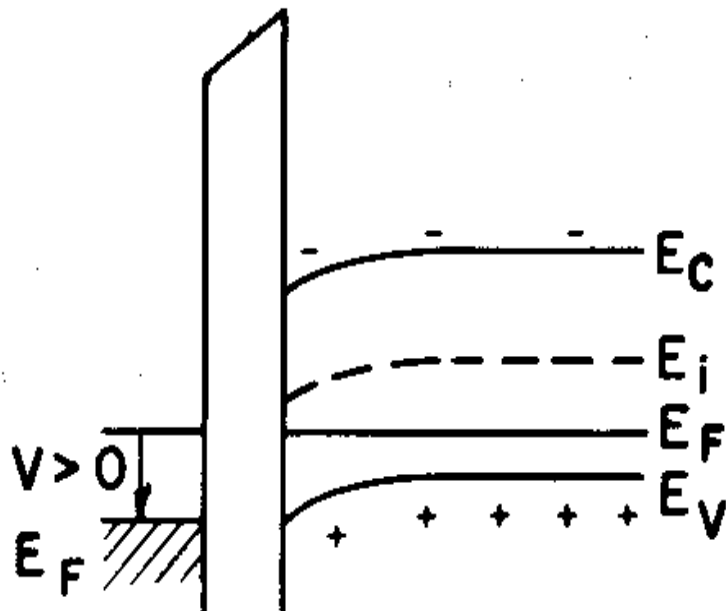
But this increases the barrier
for current flow !!



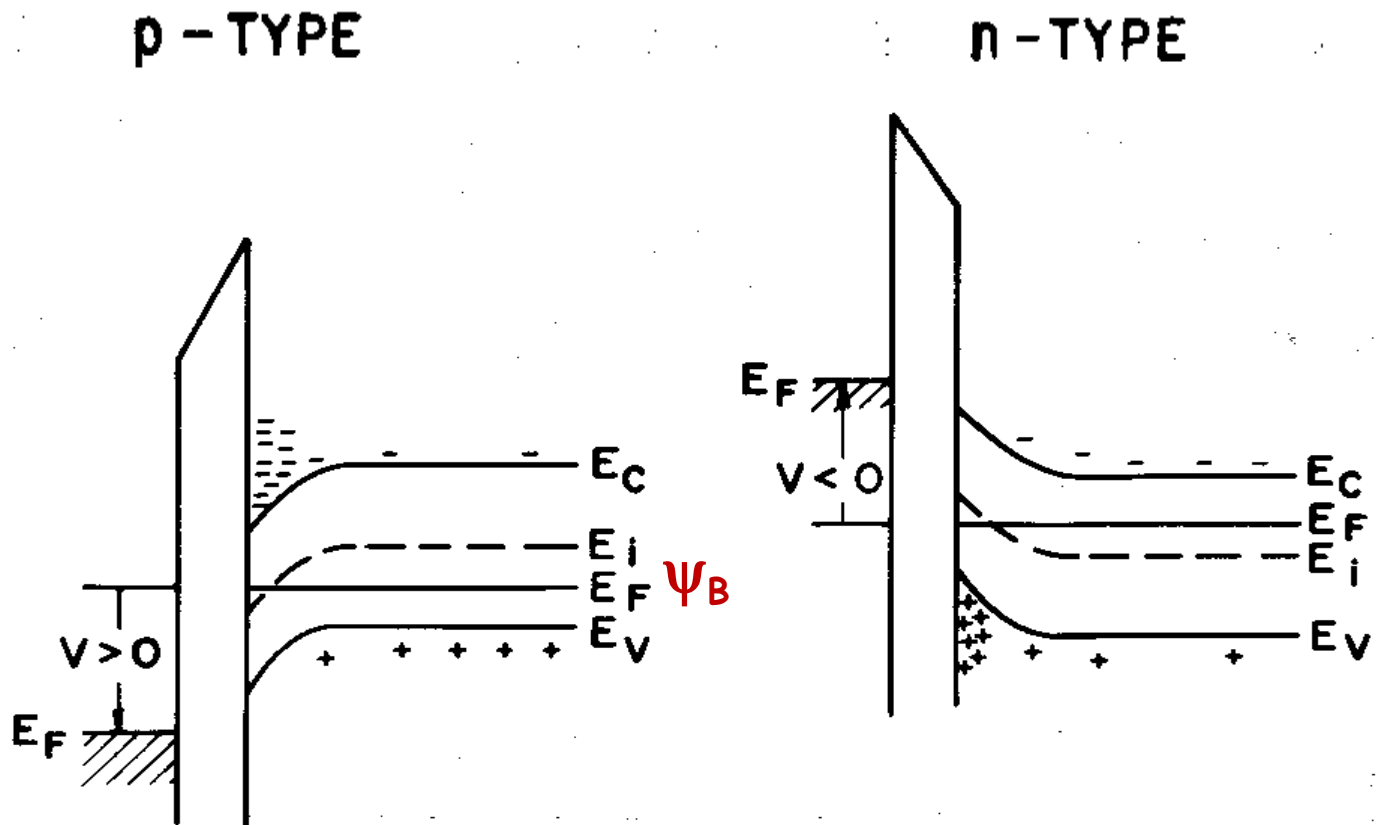
Depletion

p - TYPE

n - TYPE



Inversion



Need CB to dip below E_F .

Once below by ψ_B , minority carrier density trumps the intrinsic density.

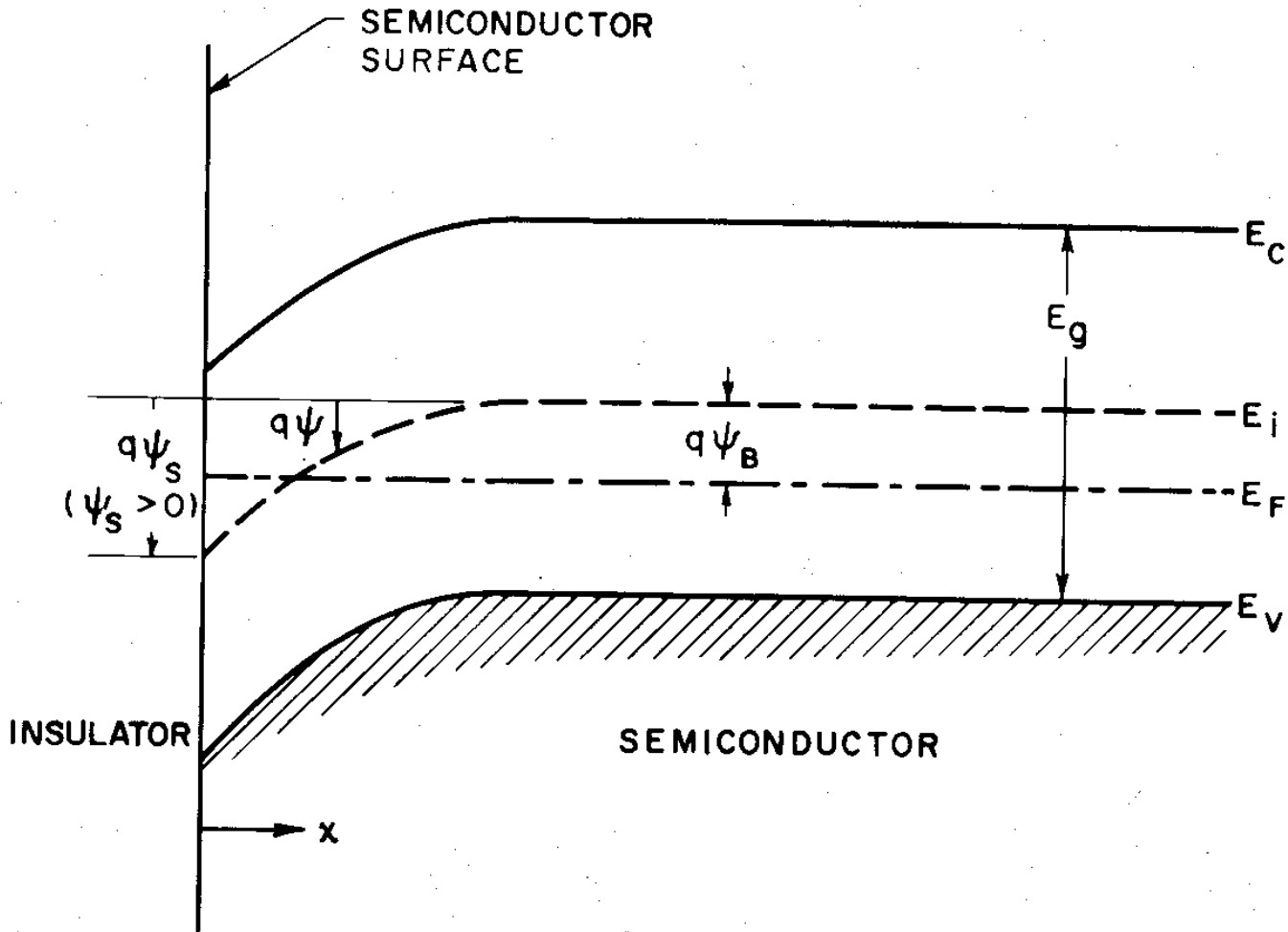
Once below by $2\psi_B$, it trumps the major carrier density (doping) !

Sometimes maths can help...



"Sure, it's an interesting concept, but do we really need mathematical proof that Casablanca should never have been colorized?"

P-type semiconductor $V_{\text{appl}} \neq 0$



Convention for p-type: ψ positive if bands bend down

Ideal MIS diode - p-type

$$n_p = n_i e^{-(E_i - E_F)/kT} = n_i e^{-(E_i - q\psi - E_F)/kT} = n_{p0} e^{q\psi/kT} = n_{p0} e^{\beta\psi}$$

CB moves towards E_F if $\psi > 0 \rightarrow n$ increases

$$p_p = p_{p0} e^{-q\psi/kT} = p_{p0} e^{-\beta\psi}$$

VB moves away from E_F if $\psi > 0 \rightarrow p$ decreases

$$\beta \equiv \frac{q}{kT}$$

Ideal MIS diode - p-type

At the semiconductor surface, $\psi = \psi_s$

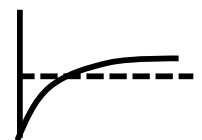
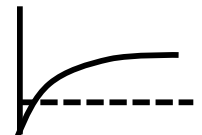
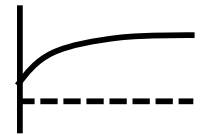
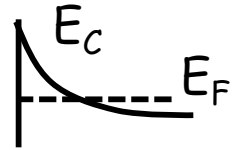
$$n_s = n_{p0} e^{\beta\psi_s}$$

$$p_s = p_{p0} e^{-\beta\psi_s}$$

Surface carrier concentration

$$n_s = n_{p0} e^{\beta\psi_s} \quad p_s = p_{p0} e^{-\beta\psi_s}$$

- $\psi_s < 0$ - accumulation of holes
- $\psi_s = 0$ - flat band
- $\psi_B > \psi_s > 0$ - depletion of holes
- $\psi_s = \psi_B$ - intrinsic concentration $n_s = p_s = n_i$
- $\psi_s > \psi_B$ - Inversion (more electrons than holes)



Want to find ψ , E-field, Capacitance

- Solve **Poisson's equation** to get E field, potential based on charge density distribution(one dimension)

$$\nabla \cdot \mathcal{E} = \rho / k\epsilon_0 = \rho / \epsilon_s = \frac{d\mathcal{E}}{dx} \rightarrow 1-D$$

$$\mathcal{E} = -\frac{d\psi}{dx}$$

$$\Rightarrow \frac{d^2\psi}{dx^2} = -\rho / \epsilon_s$$

$$\rho(x) = q(N_D^+ - N_A^- + p_p - n_p)$$

- Away from the surface, $\rho = 0$

$$\Rightarrow N_D^+ - N_A^- = n_{p0} - p_{p0}$$

- and

$$p_p - n_p = p_{p0} e^{-\beta\psi} - n_{p0} e^{\beta\psi}$$

$$\Rightarrow \frac{d^2\psi}{dx^2} = -\frac{q}{\epsilon_s} (p_{p0} (e^{-\beta\psi} - 1) - n_{p0} (e^{\beta\psi} - 1))$$

Solve Poisson's equation:

$$\Rightarrow \frac{d^2\psi}{dx^2} = -\frac{q}{\epsilon_s} (p_{p0}(e^{-\beta\psi} - 1) - n_{p0}(e^{\beta\psi} - 1))$$

$$\mathcal{E} = -d\psi/dx$$

$$\begin{aligned} d^2\psi/dx^2 &= -d\mathcal{E}/dx \\ &= (d\mathcal{E}/d\psi).(-d\psi/dx) \\ &= \mathcal{E}d\mathcal{E}/d\psi \end{aligned}$$

$$\mathcal{E}d\mathcal{E}/d\psi = -\frac{q}{\epsilon_s} (p_{p0}(e^{-\beta\psi} - 1) - n_{p0}(e^{\beta\psi} - 1))$$

Solve Poisson's equation:

- Do the integral:
- LHS:

$$\int_0^x x dx = \frac{x^2}{2} \rightarrow x = \frac{d\psi}{dx}$$

- RHS:

$$\int_0^x e^{\pm\beta x} dx, \int_0^x dx$$

- Get expression for \mathcal{E} field ($d\psi/dx$):

$$E_{field}^2 = \left(\frac{kT}{q}\right)^2 \left(\frac{qp_{p0}\beta}{2\epsilon_s}\right) \left[(e^{-\beta\psi} + \beta\psi - 1) + \frac{n_{p0}}{p_{p0}} (e^{\beta\psi} - \beta\psi - 1) \right]$$

Define:

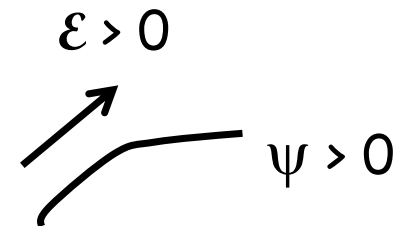
$$L_D \equiv \sqrt{\frac{kT\epsilon_s}{p_{p0}q^2}} \equiv \sqrt{\frac{\epsilon_s}{qp_{p0}\beta}} \quad \text{Debye Length}$$

$$F\left(\beta\psi, \frac{n_{p0}}{p_{p0}}\right) = \left[\left(e^{-\beta\psi} + \beta\psi - 1\right) + \frac{n_{p0}}{p_{p0}} \left(e^{\beta\psi} - \beta\psi - 1\right) \right]^{1/2}$$

Then:

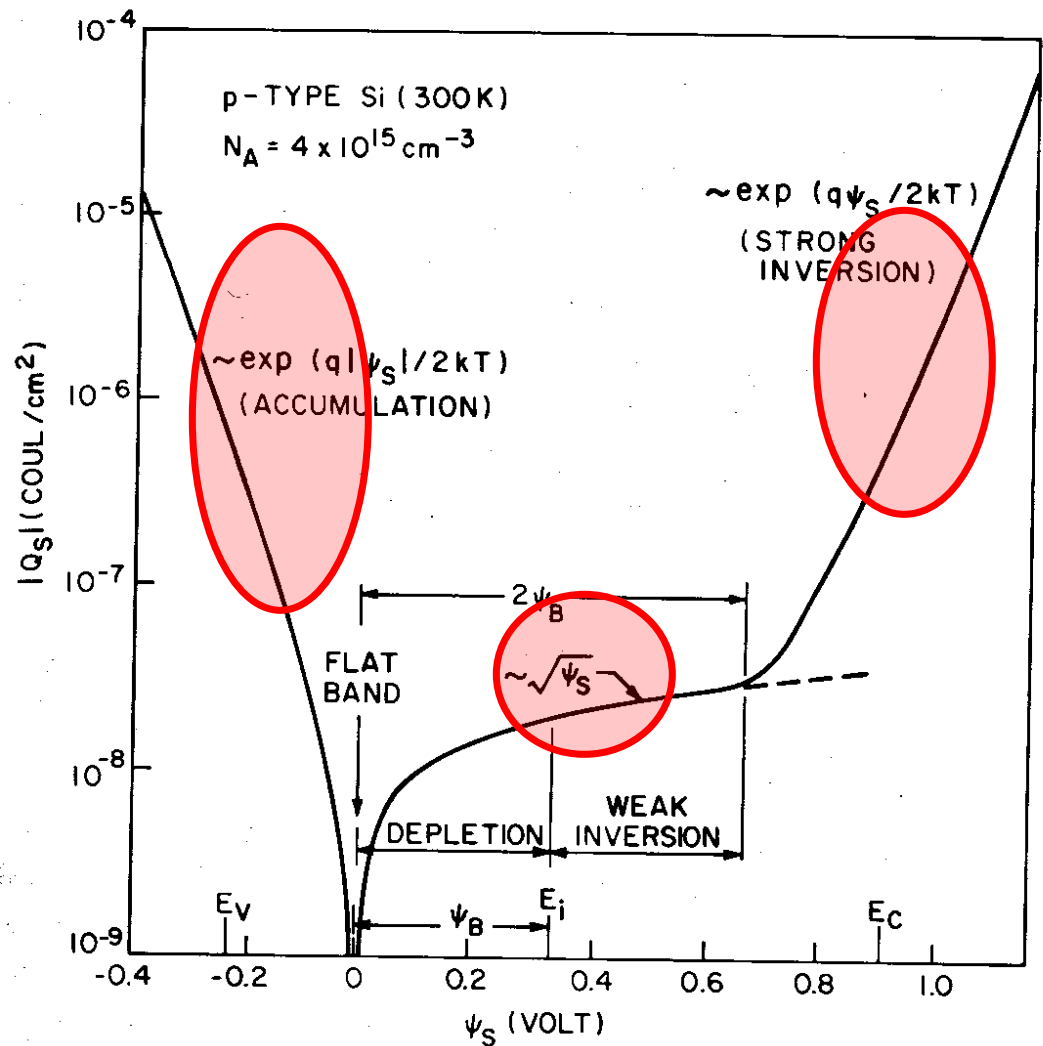
$$E_{field} = \pm \frac{\sqrt{2}kT}{qL_D} F\left(\beta\psi, \frac{n_{p0}}{p_{p0}}\right)$$

+ for $\psi > 0$ and - for $\psi < 0$



Use Gauss' Law to find surface charge per unit area

$$Q_s = -\epsilon_s E_s = \mp \frac{\sqrt{2kT}}{qL_D} F\left(\beta\psi_s, \frac{n_{p0}}{p_{p0}}\right)$$



$$Q_s = \mp \frac{\sqrt{2kT}}{qL_D} \left[\left(e^{-\beta\psi_s} - \beta\psi_s - 1 \right) + \frac{n_{p0}}{p_{p0}} \left(e^{\beta\psi_s} - \beta\psi_s - 1 \right) \right]^{1/2}$$

Accumulation to depletion to strong Inversion

- For negative ψ , first term in F dominates - exponential
- For small positive ψ , second term in F dominates - $\sqrt{\psi}$
- As ψ gets larger, $\frac{n_{p0} e^{\beta\psi}}{p_{p0}} \rightarrow 1$ second exponential gets big

$$\psi_B = (kT/q) \ln(N_A/n_i) = (1/\beta) \ln(p_{p0}/\sqrt{p_{p0}n_{p0}})$$

$$(n_{p0}/p_{p0}) = e^{-2\beta\psi_B}$$

$$\psi_S > 2\psi_B$$

Questions

✓ What is the MOS capacitance? $Q_S(\psi_S)$

✓ What are the local conditions during inversion? $\psi_{S,cr}$

How does the potential vary with position? $\psi(x)$

How much inversion charge is generated at the surface? $Q_{inv}(x, \psi_S)$

Add in the oxide: how does the voltage divide? $\psi_S(V_G), \psi_{ox}(V_G)$

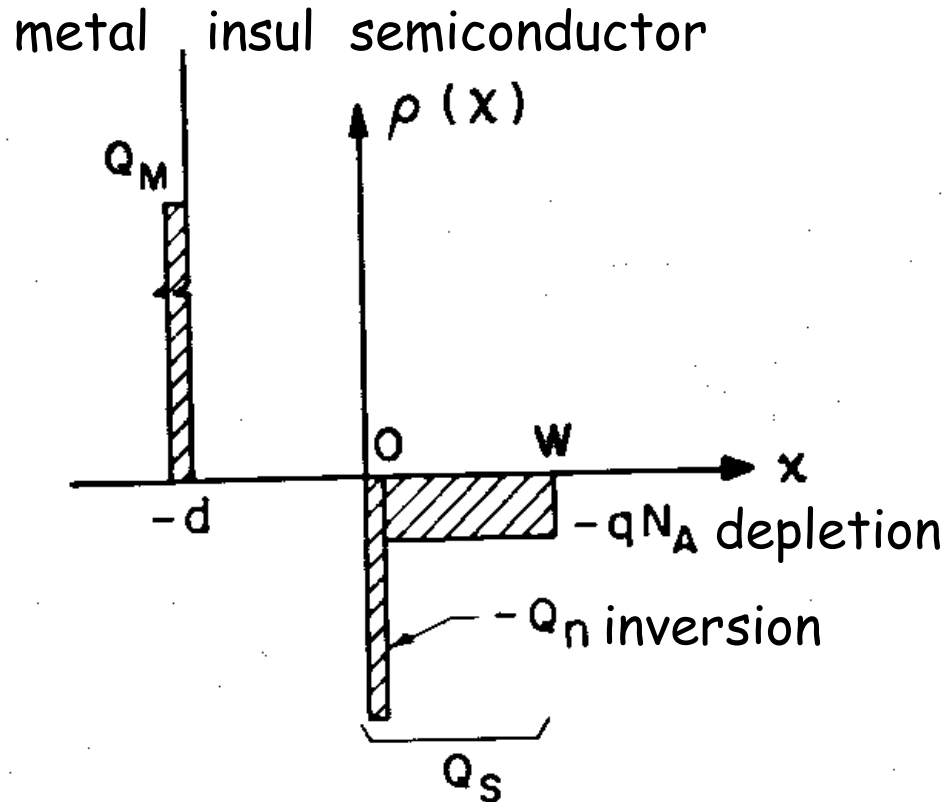
How much gate voltage do you need to invert the channel? V_{TH}

How much inversion charge is generated by the gate? $Q_{inv}(V_G)$

What's the overall C-V of the MOSFET? $Q_S(V_G)$

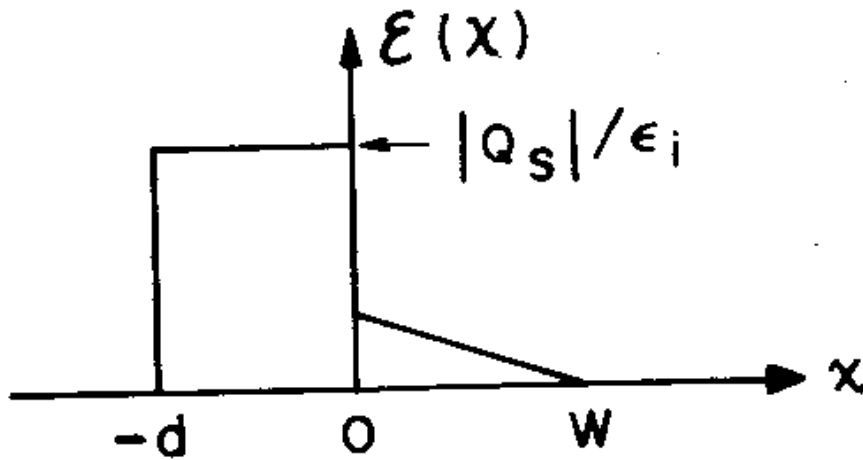
Charges, fields, and potentials

- Charge on metal = induced surface charge in semiconductor
- No charge/current in insulator (ideal)

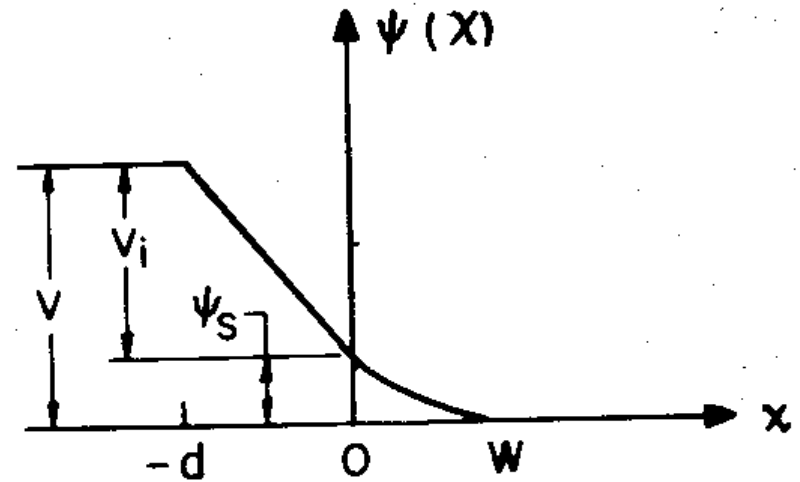


$$Q_M = Q_n + qN_A W = Q_S$$

Charges, fields, and potentials

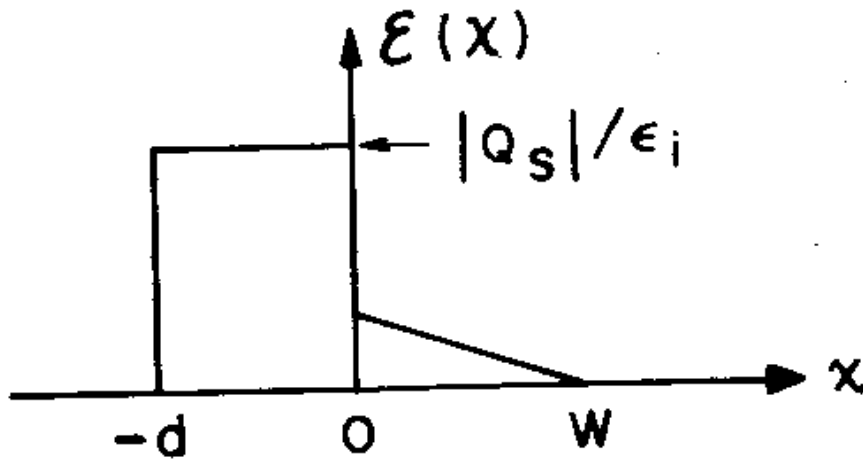


Electric Field

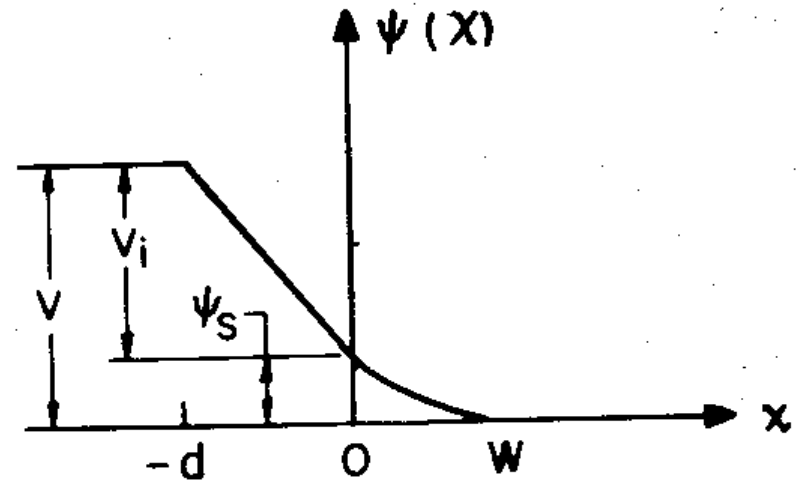


Electrostatic Potential

Depletion Region



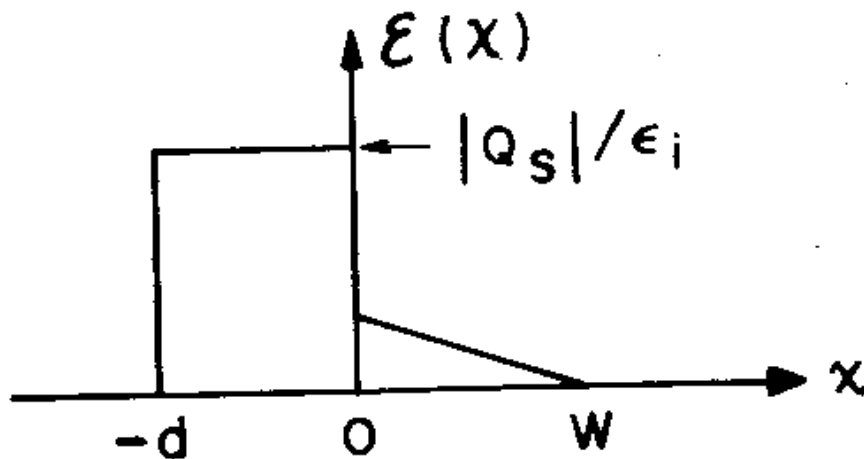
Electric Field



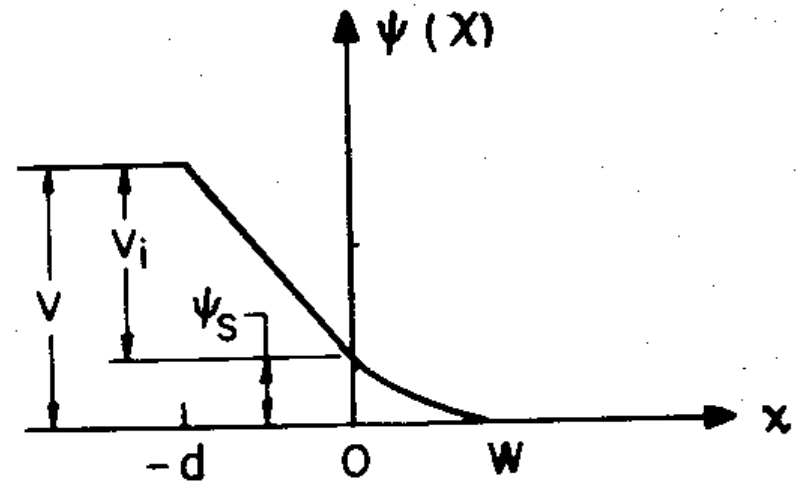
Electrostatic Potential

$$E_{field}^2 = \left(\frac{kT}{q}\right)^2 \left(\frac{qp_{p0}\beta}{2\epsilon_s}\right) \left[\cancel{e^{-\beta\psi}} + \beta\psi - 1 \right] + \frac{n_{p0}}{p_{p0}} \left(\cancel{e^{\beta\psi}} - \beta\psi - 1 \right)$$

Depletion Region



Electric Field



Electrostatic Potential

$$\psi = \psi_s(1-x/W)^2$$

$$W_{\max} = \sqrt{2\epsilon_s(2\psi_B)/qN_A}$$

$$\psi_B = (kT/q)\ln(N_A/n_i)$$

Questions

- ✓ What is the MOS capacitance? $Q_S(\psi_S)$
- ✓ What are the local conditions during inversion? $\psi_{S,cr}$
- ✓ How does the potential vary with position? $\psi(x)$

How much inversion charge is generated at the surface? $Q_{inv}(x, \psi_S)$

Add in the oxide: how does the voltage divide? $\psi_S(V_G), \psi_{ox}(V_G)$

How much gate voltage do you need to invert the channel? V_{TH}

How much inversion charge is generated by the gate? $Q_{inv}(V_G)$

What's the overall C-V of the MOSFET? $Q_S(V_G)$

Couldn't we just solve
this exactly?



Exact Solution

$$U = \beta\psi$$

$$U_S = \beta\psi_S$$

$$U_B = \beta\psi_B$$

$$d\psi/dx = -(\sqrt{2kT/qL_D})F(\psi_B, n_{p0}/p_{p0})$$

$$\int_{U_S}^U dU/F(U) = \pm x/L_D$$

$$F(U) = [e^{U_B}(e^{-U}-1+U)-e^{-U_B}(e^U-1-U)]^{1/2}$$

Exact Solution

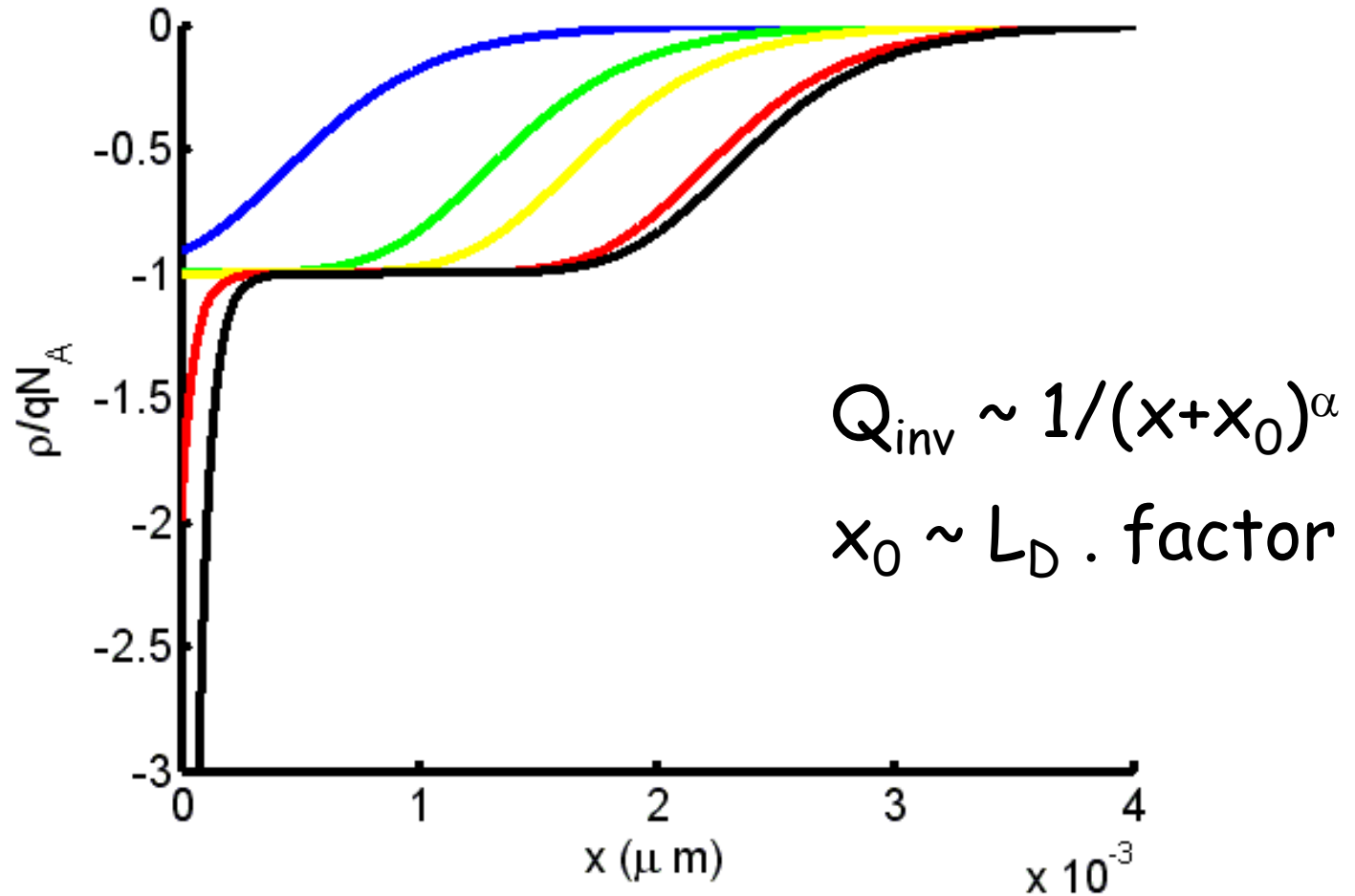
$$\rho = qn_i[e^{U_B}(e^{-U}-1) - e^{-U_B}(e^U-1)]$$

$$\int_U^{U_s} dU'/F(U',U_B) = \pm x/L_D$$

$$F(U,U_B) = [e^{U_B}(e^{-U}-1+U) + e^{-U_B}(e^U-1-U)]^{1/2}$$

Exact Solution

$$N_A = 1.67 \times 10^{15}$$



Questions

- ✓ What is the MOS capacitance? $Q_S(\psi_S)$
- ✓ What are the local conditions during inversion? $\psi_{S,cr}$
- ✓ How does the potential vary with position? $\psi(x)$
- ✓ How much inversion charge is generated at the surface? $Q_{inv}(x, \psi_S)$

Add in the oxide: how does the voltage divide? $\psi_S(V_G)$, $\psi_{ox}(V_G)$

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What's the overall C-V of the MOSFET? $Q_S(V_G)$

Threshold Voltage for Strong Inversion

- Total voltage across MOS structure= voltage across dielectric plus ψ_s

$$V_T(\text{strong_inversion}) = V_i + \psi_s = \frac{Q_s}{C_i} + 2\psi_B$$

$$Q_s(SI) = qN_A W_{\max} = qN_A \sqrt{\frac{2\epsilon_s \psi_s(\text{inv})}{qN_A}} = \sqrt{2\epsilon_s qN_A (2\psi_B)}$$

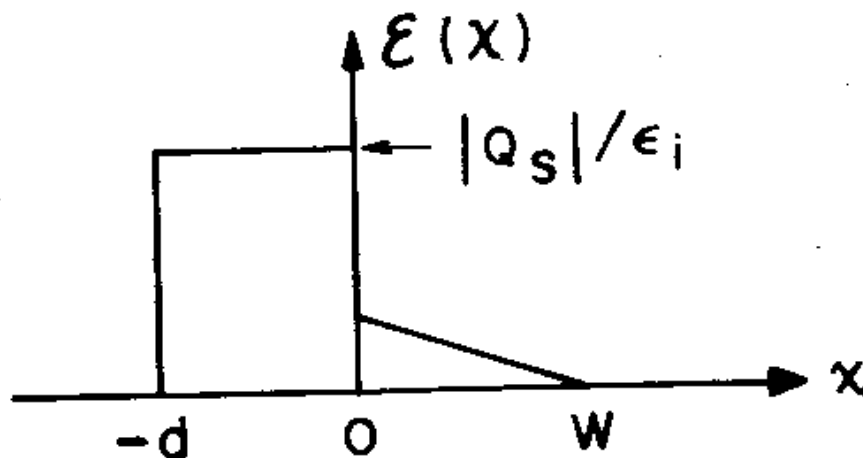
$$\Rightarrow V_T = \frac{\sqrt{2\epsilon_s qN_A (2\psi_B)}}{C_i} + 2\psi_B$$

Notice Boundary Condition !!

$$\epsilon_{ox} V_i / t_{ox} = \epsilon_s \psi_s / (W/2) \text{ Before Inversion}$$

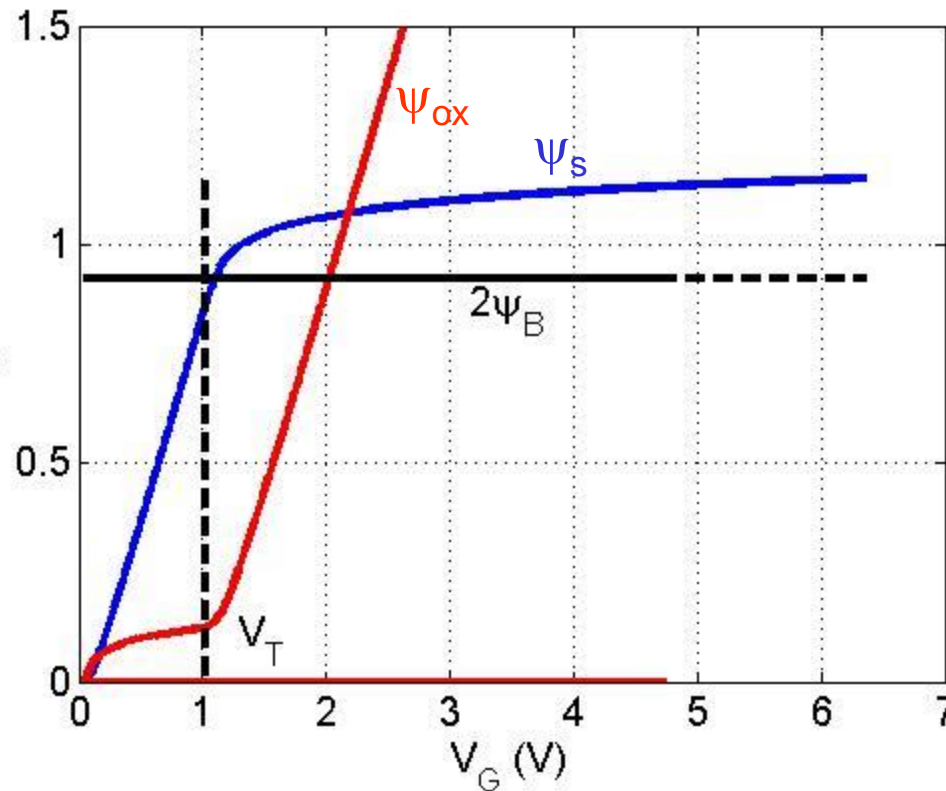
After inversion there is a discontinuity in D due to surface Q_{inv}

$$\begin{aligned} V_{ox} (\text{at threshold}) &= \epsilon_s (2\psi_B) / (W_{max}/2) C_i = \\ &= \frac{\sqrt{2\epsilon_s q N_A (2\psi_B)}}{C_i} \end{aligned}$$



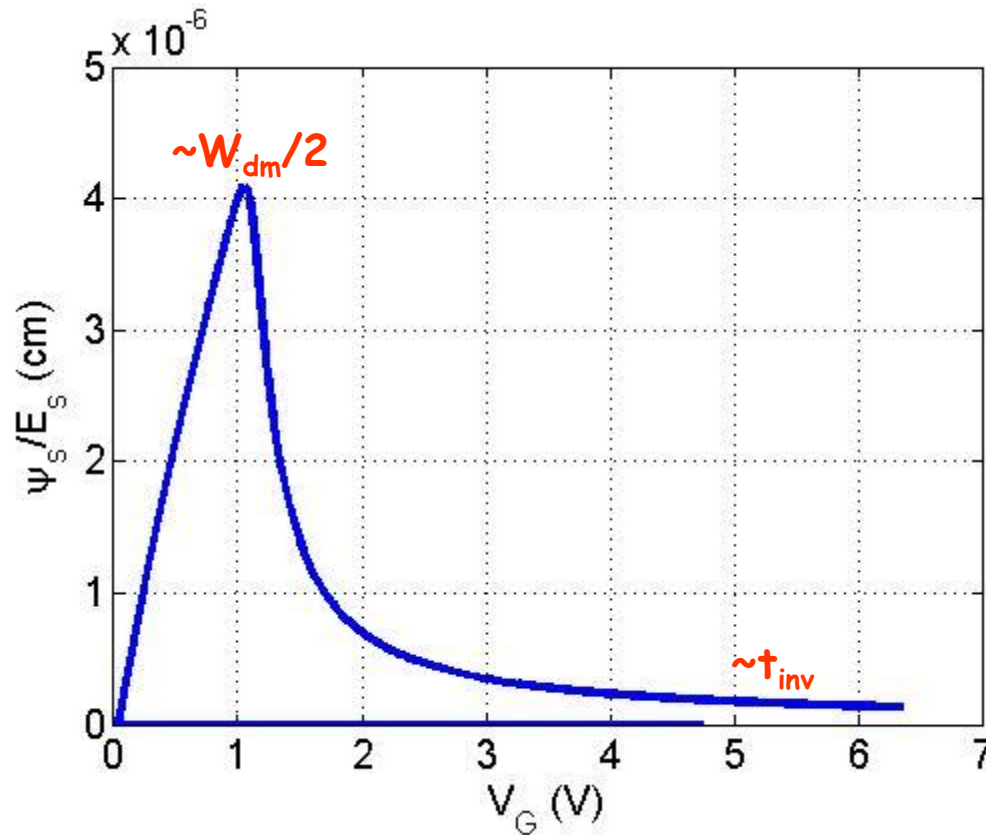
Local Potential vs Gate voltage

$$V_G = V_{fb} + \psi_s + (\kappa_s t_{ox} / \kappa_{ox}) \sqrt{(2kTN_A / \epsilon_0 \kappa_s) [\beta \psi_s + e^{\beta(\psi_s - 2\psi_B)}]^{1/2}}$$



Initially, all voltage drops across channel (blue curve). Above threshold, channel potential stays pinned to $2\psi_B$, varying only logarithmically, so that most of the gate voltage drops across the oxide (red curve).

Look at Effective charge width



Initially, a fast increasing channel potential drops across increasing depletion width

Eventually, a constant potential drops across a decreasing inversion layer width, so field keeps increasing and thus matches increasing field in oxide

Questions

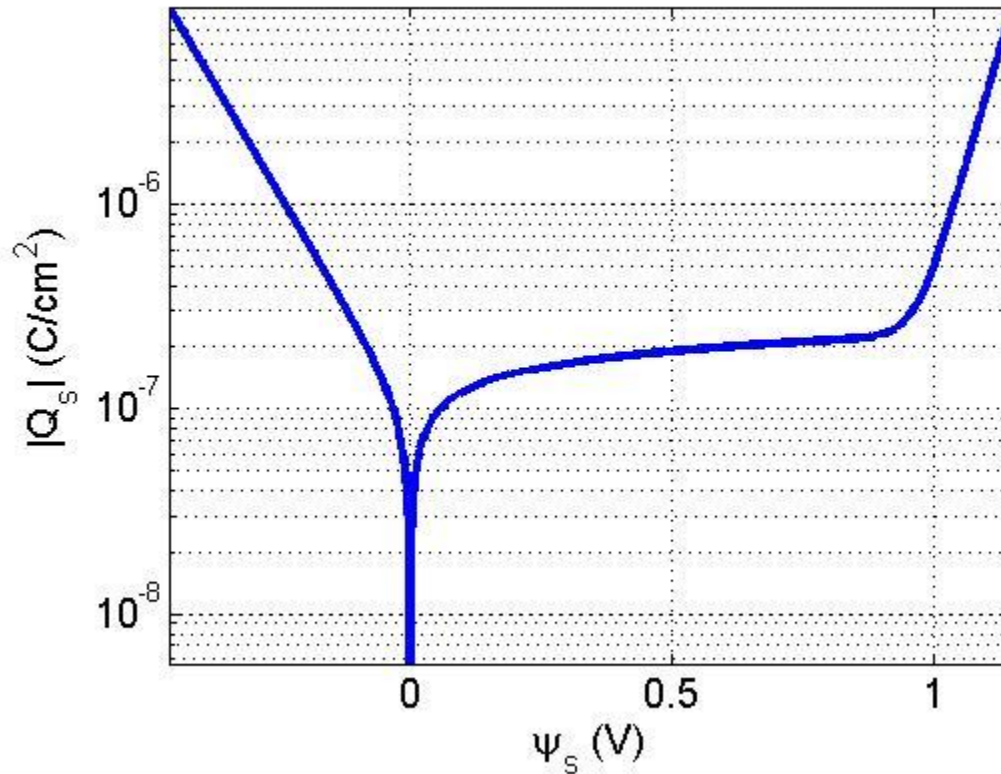
- ✓ What is the MOS capacitance? $Q_S(\psi_S)$
- ✓ What are the local conditions during inversion? $\psi_{S,cr}$
- ✓ How does the potential vary with position? $\psi(x)$
- ✓ How much inversion charge is generated at the surface? $Q_{inv}(x, \psi_S)$
- ✓ Add in the oxide: how does the voltage divide? $\psi_S(V_G), \psi_{ox}(V_G)$
- ✓ How much gate voltage do you need to invert the channel? V_{TH}

How much inversion charge is generated by the gate? $Q_{inv}(V_G)$

What's the overall C-V of the MOSFET? $Q_S(V_G)$

Charge vs Local Potential

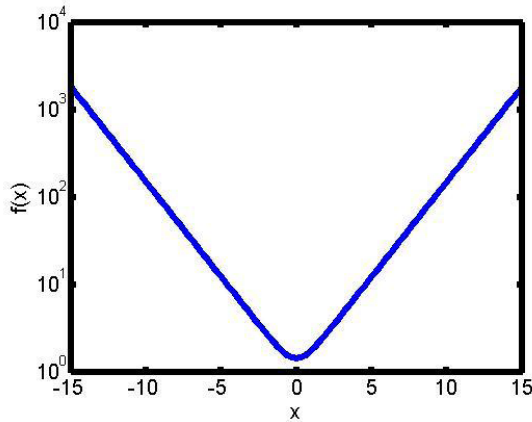
$$Q_s \approx \sqrt{(2\epsilon_0\kappa_s k T N_A)[\beta\psi_s + e^{\beta(\psi_s - 2\psi_B)}]^{1/2}}$$



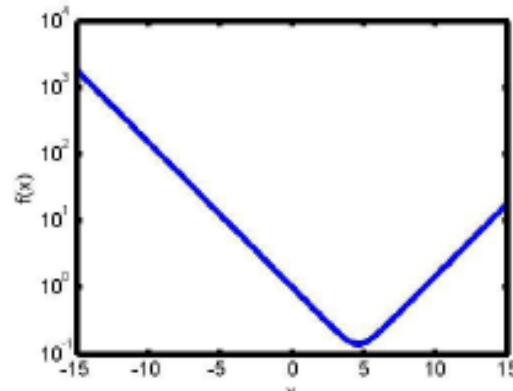
Beyond threshold, all charge goes to inversion layer

NEW

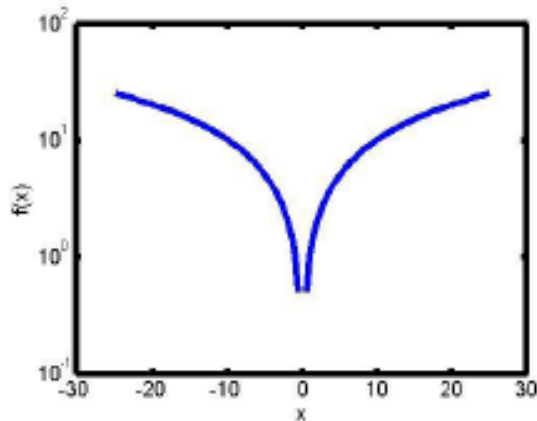
How do we get the curvatures?



$[\exp(-\beta x) + n_r \cdot \exp(\beta x)]^{1/2}$ with $n_r = 1$.

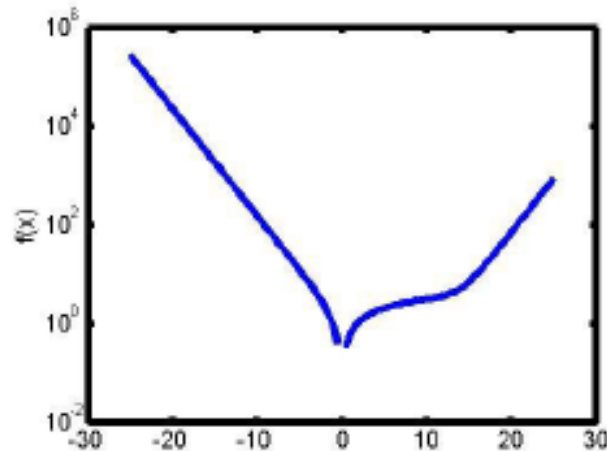


$n_r = 10^{-3}$.



Add other terms and keep
Leading term

$$(x^2 + n_r \cdot [x^2])^{1/2}$$



EXACT

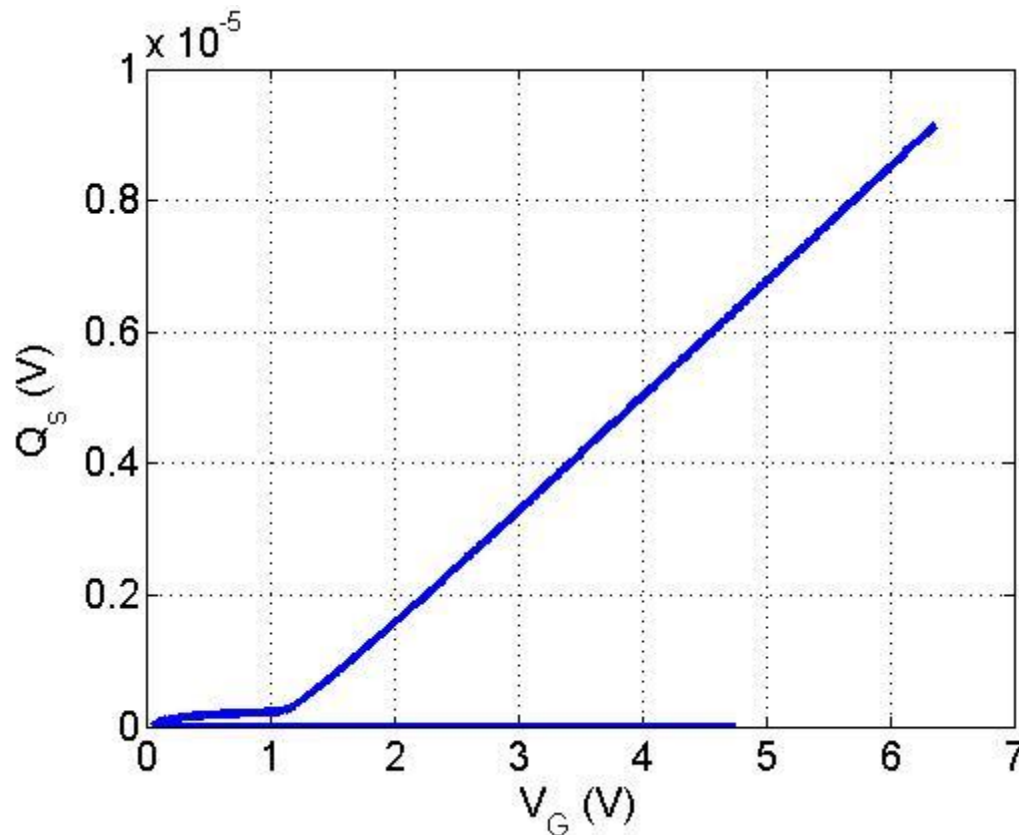
$$([\exp(-\beta x) + \beta x - 1] + n_r \cdot [\exp(\beta x) - \beta x - 1])^{1/2}$$

Inversion Charge vs Gate voltage

$$Q \sim e^{\beta(\psi_s - 2\psi_B)}, \psi_s - 2\psi_B \sim \log(V_G - V_T)$$

Exponent of a logarithm gives a linear variation of Q_{inv} with V_G

$$Q_{inv} = -C_{ox}(V_G - V_T)$$



Why C_{ox} ?

Questions

- ✓ What is the MOS capacitance? $Q_S(\psi_S)$
 - ✓ What are the local conditions during inversion? $\psi_{S,cr}$
 - ✓ How does the potential vary with position? $\psi(x)$
 - ✓ How much inversion charge is generated at the surface? $Q_{inv}(x, \psi_S)$
 - ✓ Add in the oxide: how does the voltage divide? $\psi_S(V_G), \psi_{ox}(V_G)$
 - ✓ How much gate voltage do you need to invert the channel? V_{TH}
 - ✓ How much inversion charge is generated by the gate? $Q_{inv}(V_G)$
- What's the overall C-V of the MOSFET? $Q_S(V_G)$

Capacitance

$$C_D = \frac{\partial Q_S}{\partial \psi} = \frac{\epsilon_S}{\sqrt{2}L_D} \frac{\left[1 - e^{-\beta\psi_s} + \left(\frac{n_{p0}}{p_{p0}} \right) (e^{-\beta\psi_s} - 1) \right]}{F\left(\beta\psi_s, \frac{n_{p0}}{p_{p0}} \right)}$$

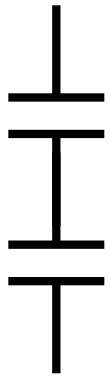
For $\psi_s=0$ (Flat Band):

Expand exponentials..... $e^x = 1 + x + \frac{x^2}{2!} + \frac{x^3}{3!} + \dots$

$$C_D(\text{flat_band}) = \frac{\epsilon_S}{L_D}$$

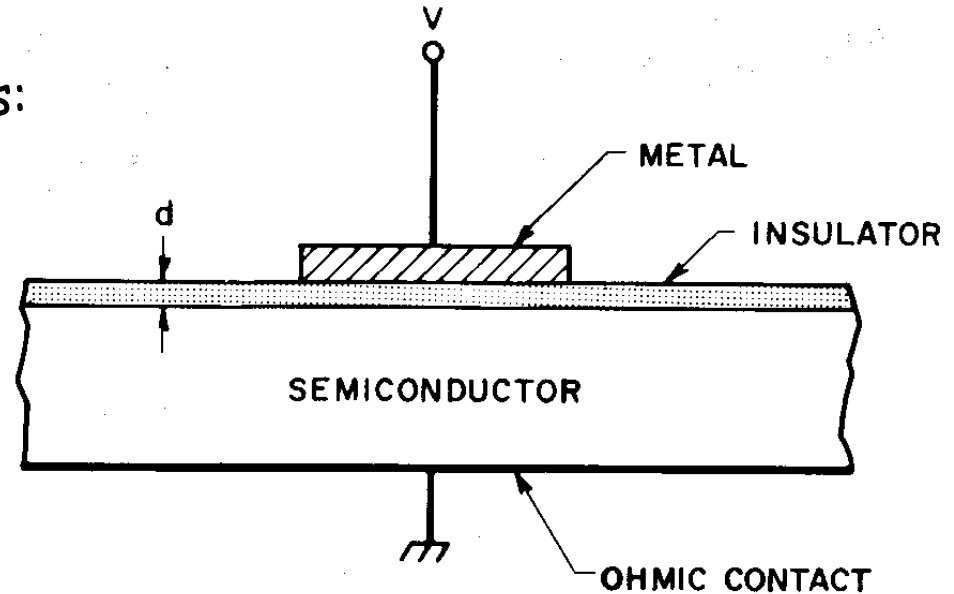
Capacitance of whole structure

- Two capacitors in series:



C_i - insulator

C_D - Depletion



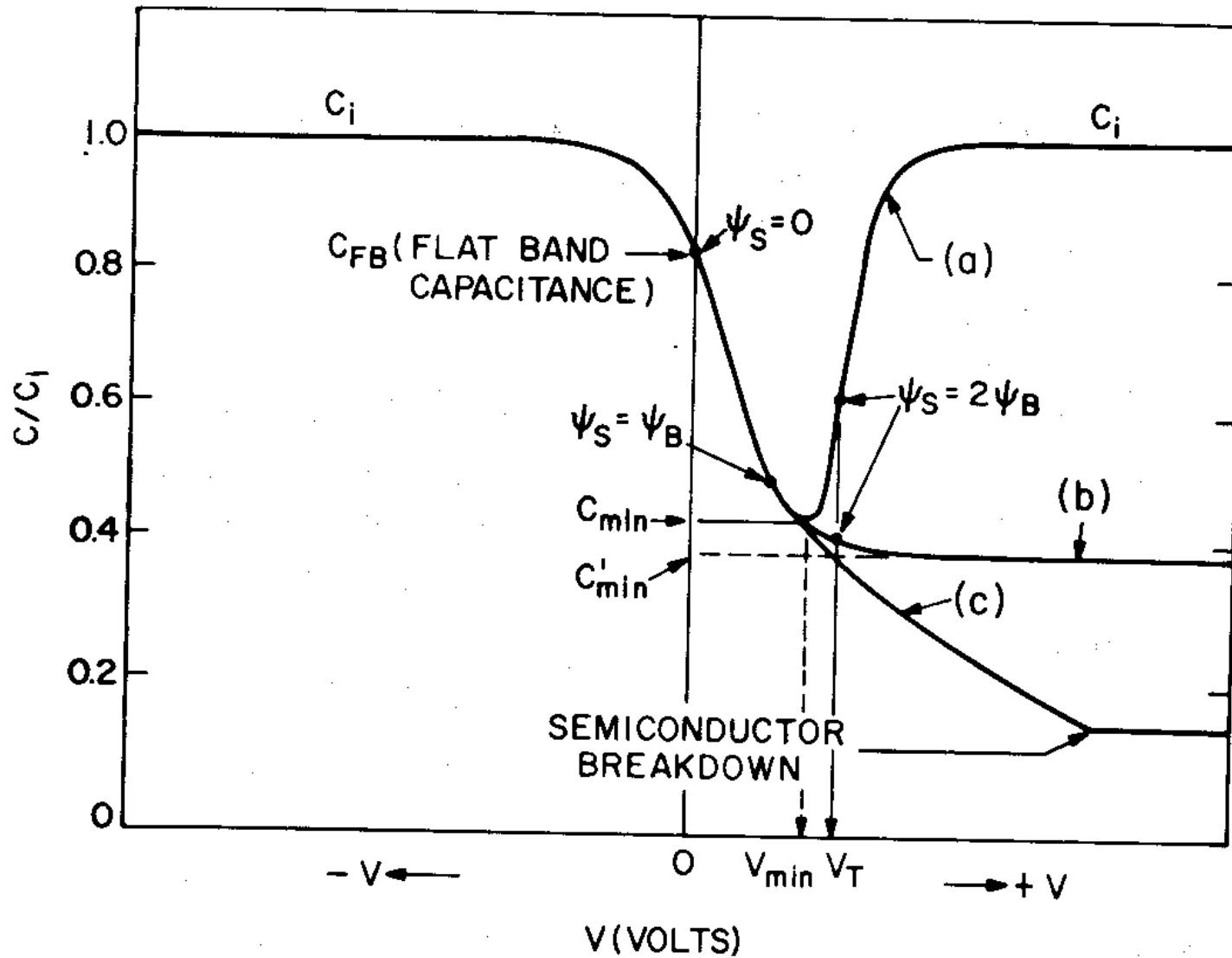
$$\frac{1}{C} = \frac{1}{C_i} + \frac{1}{C_D}$$

OR

$$C = \frac{C_i C_D}{C_i + C_D}$$

$$C_i \equiv \frac{\epsilon_i}{d}$$

Capacitance vs Voltage



Flat Band Capacitance

- Negative voltage = accumulation - $C \sim C_i$
- Zero voltage - Flat Band

$$V = 0 \Rightarrow \psi = 0 \Rightarrow C = C_{FB}$$

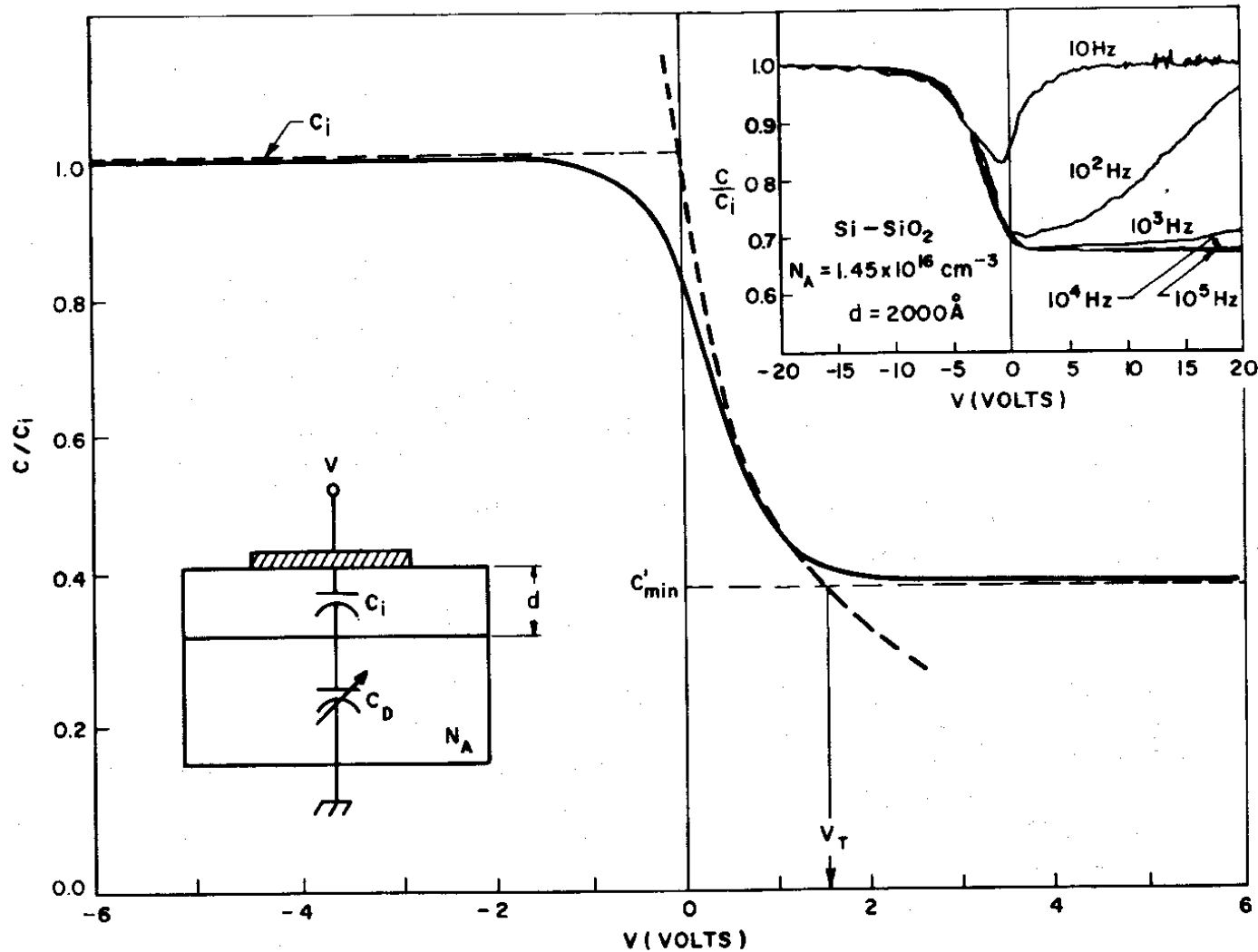
$$\frac{1}{C_{FB}} = \frac{1}{C_i} + \frac{1}{C_D} = \frac{1}{\frac{\epsilon_i}{d}} + \frac{1}{\frac{\epsilon_s}{L_D}} = \frac{\epsilon_s d + \epsilon_i L_D}{\epsilon_i \epsilon_s} = \frac{d + \frac{\epsilon_i}{\epsilon_s} L_D}{\epsilon_i}$$

$$\Rightarrow C_{FB} = \frac{\epsilon_i}{d + \frac{\epsilon_i}{\epsilon_s} L_D}$$

CV

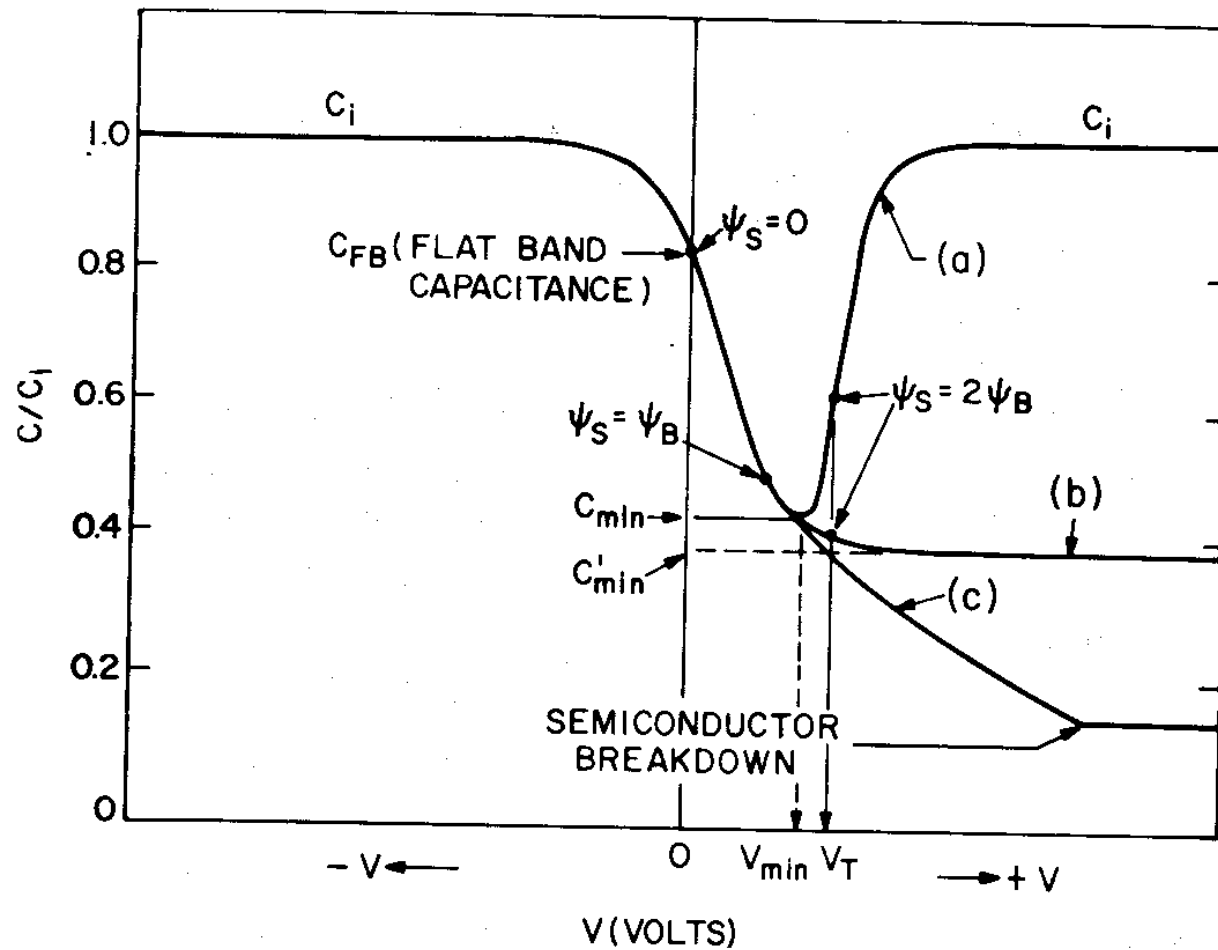
- As voltage is increased, C goes through minimum (weak inversion) where $d\psi/dQ$ is fairly flat
- C will increase with onset of strong inversion
- Capacitance is an AC measurement
- Only increases when AC period long wrt minority carrier lifetime
- At "high" frequency, carriers can't keep up - don't see increased capacitance with voltage
- For Si MOS, "high" frequency = 10-100 Hz

CV Curves - Ideal MOS Capacitor

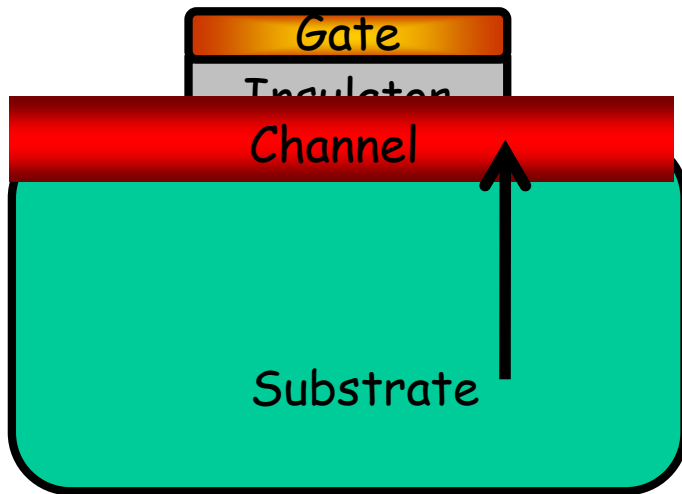


$$C'_{min} = \frac{\epsilon_i}{d + \frac{\epsilon_i}{\epsilon_s} W_{max}}$$

MOScap vs MOSFET

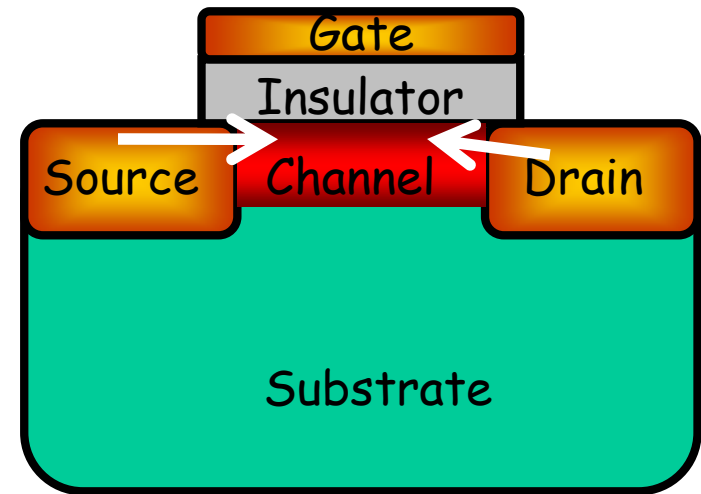


MOScap vs MOSFET



Minority carriers generated by
RG, over minority carrier lifetime
 $\sim 100\mu\text{s}$

So C_{inv} can be $\ll C_{\text{ox}}$ if fast gate
switching ($\sim \text{GHz}$)



Majority carriers pulled in
from contacts (fast !!)

$$C_{\text{inv}} = C_{\text{ox}}$$

Example Metal-SiO₂-Si

- $N_A = 10^{17}/\text{cm}^3$
- At room temp $kT/q = 0.026\text{V}$
- $n_i = 9.65 \times 10^9/\text{cm}^3$
- $\epsilon_s = 11.9 \times 1.85 \times 10^{-14} \text{ F/cm}$

$$W_{\max} = \sqrt{\frac{4\epsilon_s kT \ln\left(\frac{N_A}{n_i}\right)}{q^2 N_A}} = \sqrt{\frac{11.9 \times 8.85 \times 10^{-14} \times 0.026 \ln(10^{17}/9.65 \times 10^9)}{1.6 \times 10^{-19} \times 10^{17}}}$$
$$W_{\max} = 10^{-5} \text{ cm} = 0.1 \mu\text{m}$$

Example Metal-SiO₂-Si

- $d=50 \text{ nm thick oxide}=10^{-5} \text{ cm}$
- $\epsilon_i=3.9 \times 8.85 \times 10^{-14} \text{ F/cm}$

$$C_i = \frac{\epsilon_i}{d} = \frac{3.9 \times 8.85 \times 10^{-14}}{10^{-5}} = 6.9 \times 10^{-7} \text{ F/cm}^2$$

$$\psi_s(inv) = 2\psi_B = \frac{2kT}{q} \ln\left(\frac{N_A}{n_i}\right) = 2 \times 0.026 \times \ln\left(\frac{10^{17}}{9.65 \times 10^9}\right) = 0.84 \text{ Volts}$$

$$C'_{\min} = \frac{\epsilon_i}{d + \frac{\epsilon_i}{\epsilon_s} W_{\max}} = \frac{3.9 \times 8.85 \times 10^{-14}}{5 \times 10^{-7} + (3.9/11.9)10^{-5}} = 9.1 \times 10^{-8} \text{ F/cm}^2$$

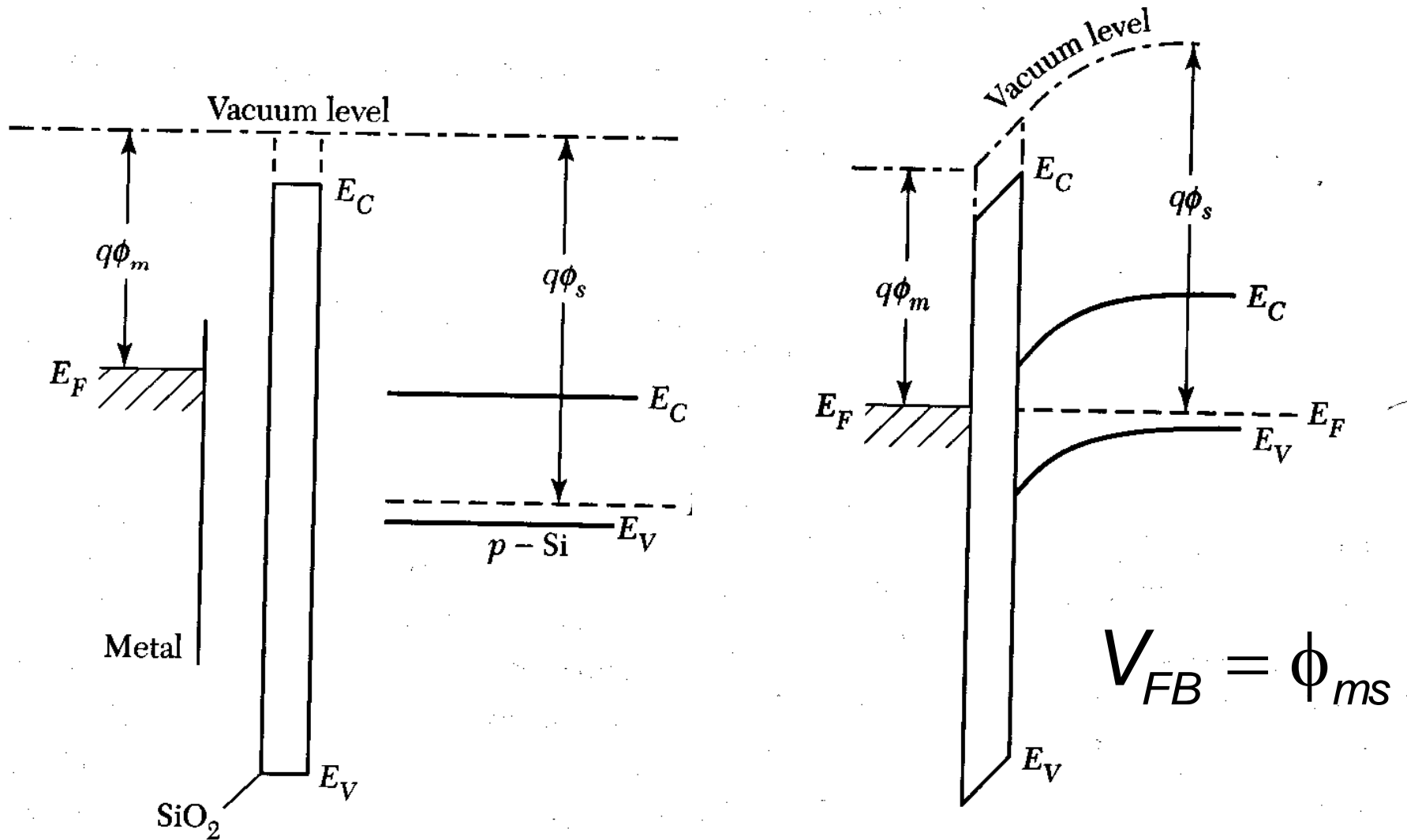
$$\frac{C'_{\min}}{C_i} = 0.13$$

$$V_{TH} = \frac{qN_A W_{\max}}{C_i} + 2\psi_B = \frac{1.6 \times 10^{-19} \times 10^{17} \times 10^{-5}}{6.9 \times 10^{-7}} + \psi_s(inv) = 0.23 + 0.84 = 1.07 \text{ Volts}$$

Real MIS Diode: Metal(poly)-Si-SiO₂ MOS

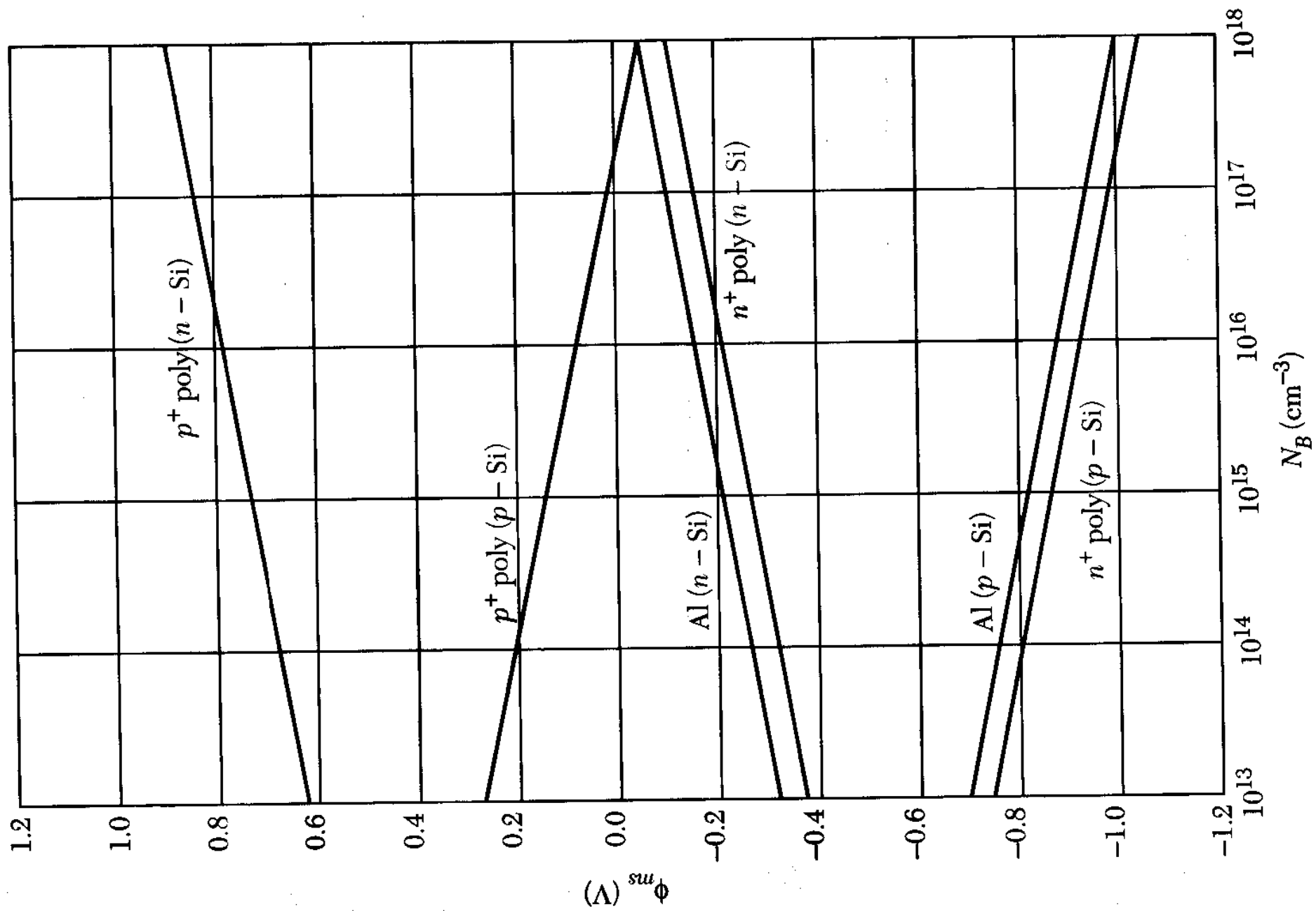
- Work functions of gate and semiconductor are NOT the same
- Oxides are not perfect
 - Trapped, interface, mobile charges
 - Tunneling
- All of these will effect the CV characteristic and threshold voltage

Band bending due to work function difference

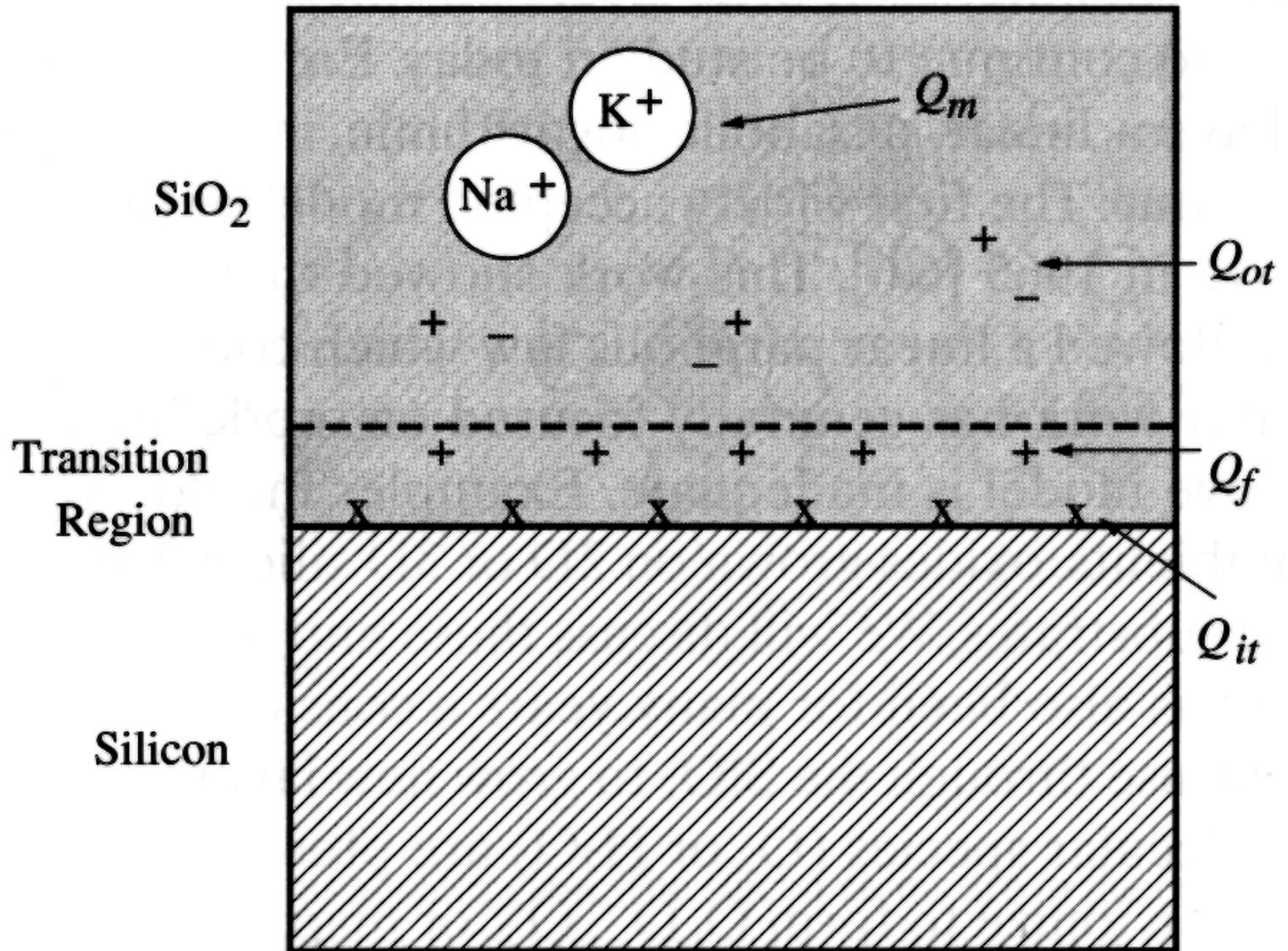


Work Function Difference

- $q\phi_s$ = semiconductor work function = difference between vacuum and Fermi level
- $q\phi_m$ = metal work function
- $q\phi_{ms} = (q\phi_m - q\phi_s)$
- For Al, $q\phi_m = 4.1 \text{ eV}$
- n^+ polysilicon $q\phi_s = 4.05 \text{ eV}$
- p^+ polysilicon $q\phi_s = 5.05 \text{ eV}$
- $q\phi_{ms}$ varies over a wide range depending on doping



SiO_2 -Si Interface Charges



Standard nomenclature for Oxide charges:

Q_M = Mobile charges (Na⁺/K⁺) - can cause unstable threshold shifts - cleanliness has eliminated this issue

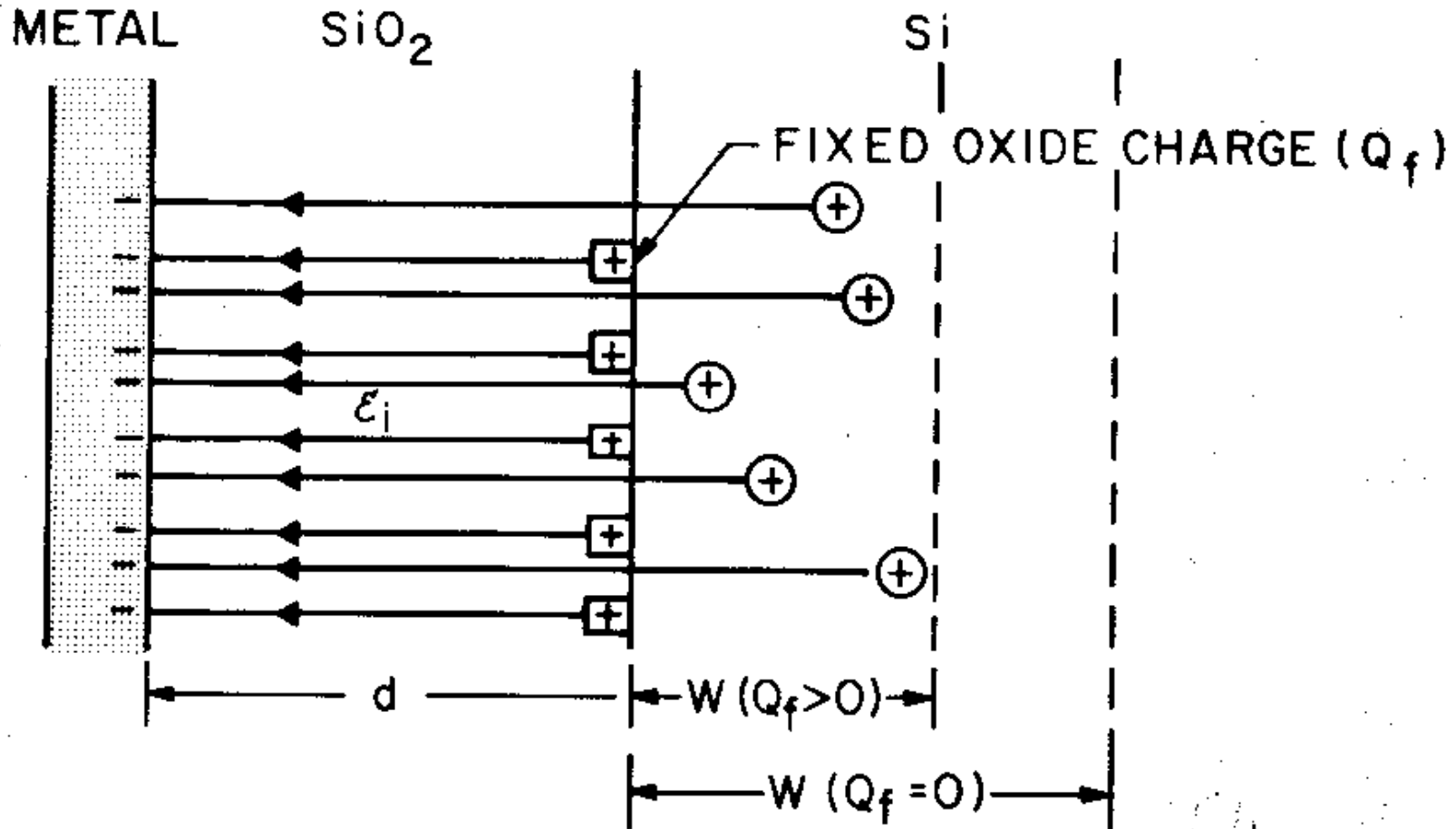
Q_{OT} = Oxide trapped charge - Can be anywhere in the oxide layer. Caused by broken Si-O bonds - caused by radiation damage e.g. alpha particles, plasma processes, hot carriers, EPROM

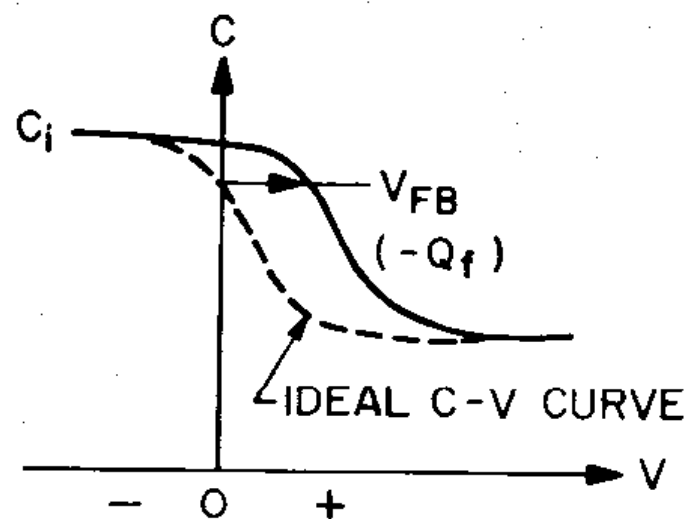
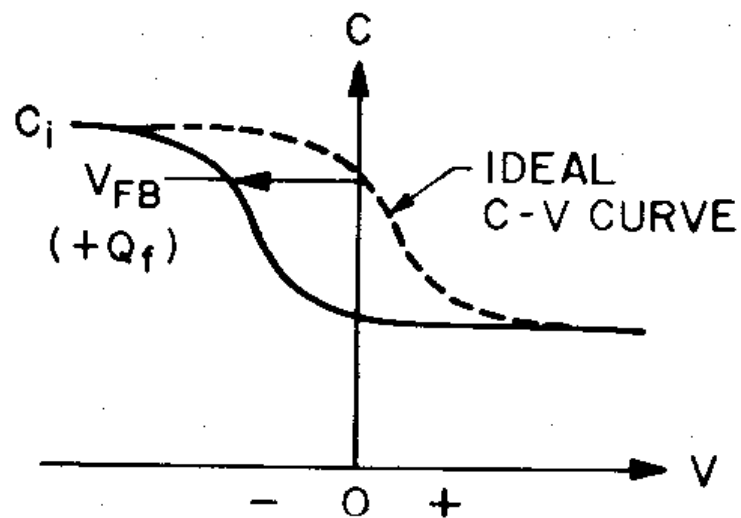
Q_F = Fixed oxide charge - positive charge layer near ($\sim 2\text{nm}$) Caused by incomplete oxidation of Si atoms (dangling bonds)
Does not change with applied voltage

Q_{IT} = Interface trapped charge. Similar in origin to Q_F but at interface. Can be pos, neg, or neutral. Traps e^- and h during device operation. Density of Q_{IT} and Q_F usually correlated - similar mechanisms. Cure is H_2 anneal at the end of the process.

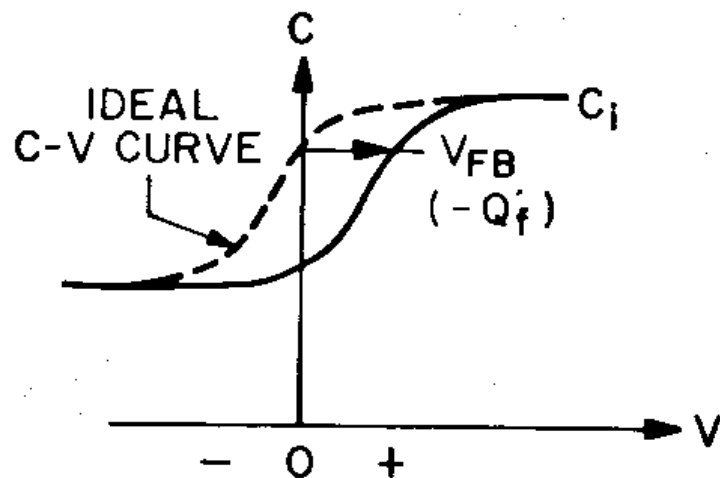
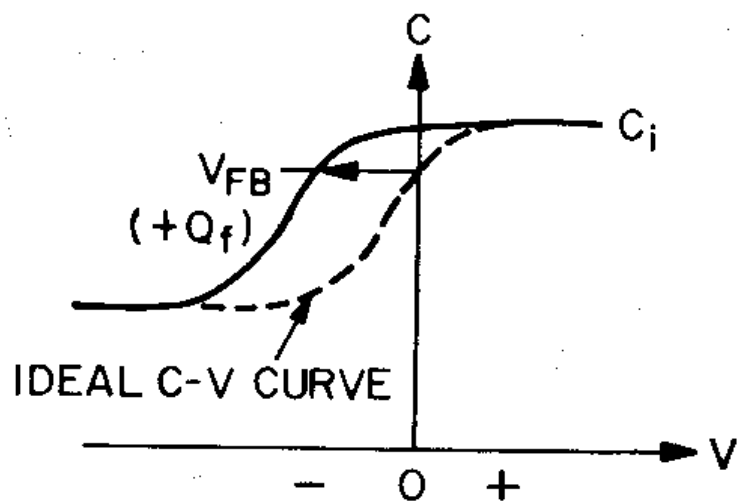
Oxide charges measured with $C-V$ methods

Effect of Fixed Oxide Charges

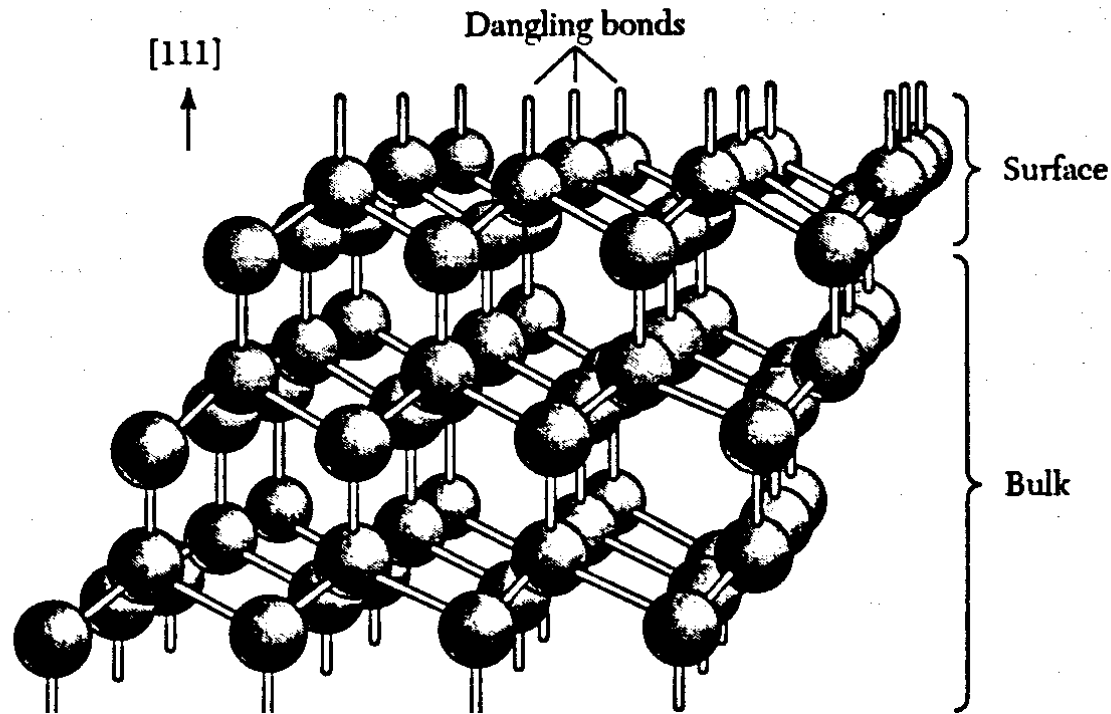




(a)



Surface Recombination



Lattice periodicity broken at surface/interface - mid-gap E levels
Carriers generated-recombined per unit area

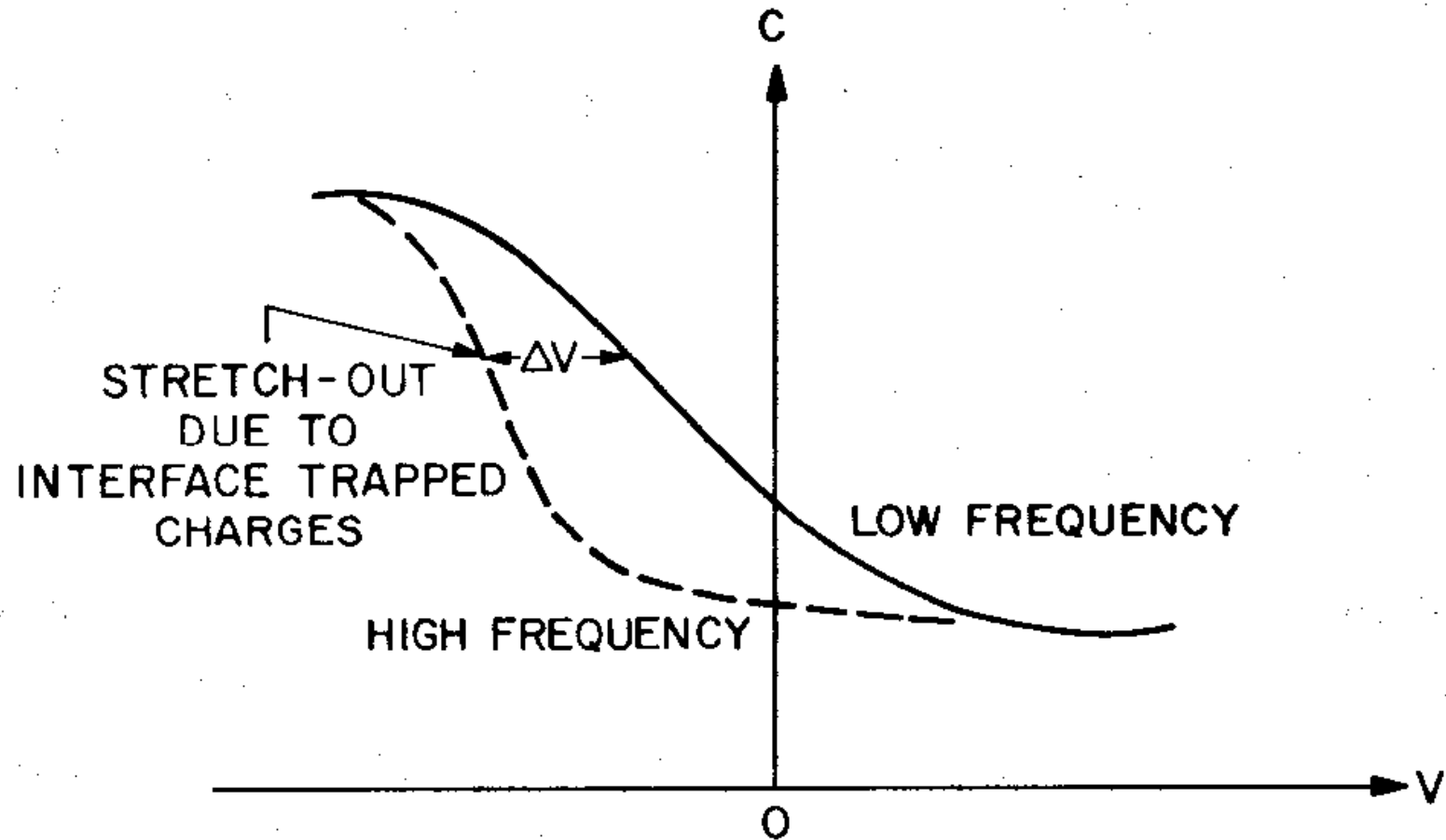
Interface Trapped Charge - Q_{IT}

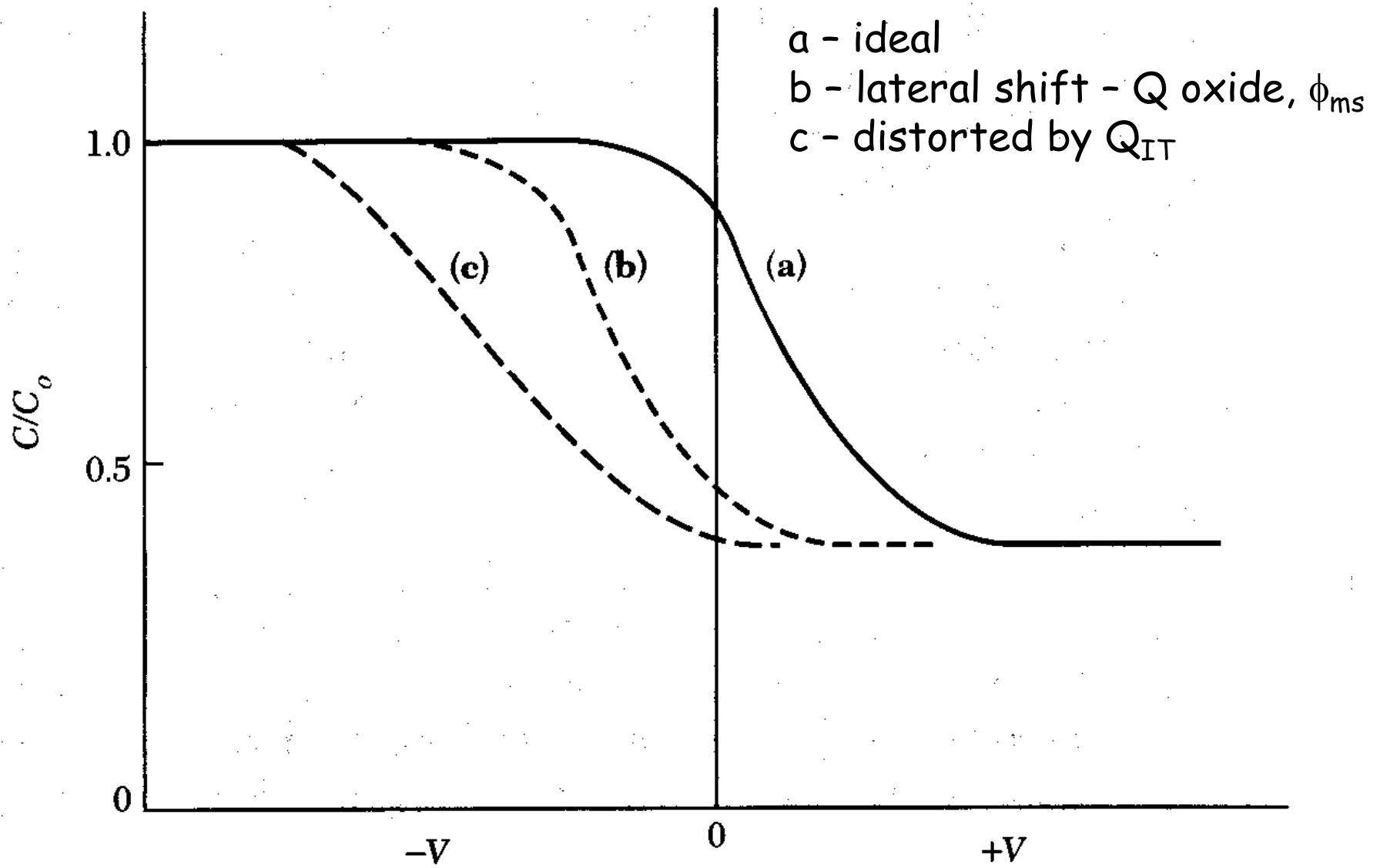
- Surface states - R-G centers caused by disruption of lattice periodicity at surface
- Trap levels distributed in band gap, with Fermi-type distributed:

$$\frac{N_D^+}{N_D} = \frac{1}{1 + g_D e^{-(E_F - E_D)/kT}}$$

- Ionization and polarity will depend on applied voltage (above or below Fermi level)
- Frequency dependent capacitance due to surface recombination lifetime compared with measurement frequency
- Effect is to distort CV curve depending on frequency
- Can be passivated w/H anneal - $10^{10}/\text{cm}^2$ in Si/SiO₂ system

Effect of Interface trapped charge on C-V curve





Non-Ideal MOS capacitor C-V curves

- Work function difference and oxide charges shift CV curve in voltage from ideal case
- CV shift changes threshold voltage
- Mobile ionic charges can change threshold voltage as a function of time - reliability problems
- Interface Trapped Charge distorts CV curve - frequency dependent capacitance
- Interface state density can be reduced by H annealing in Si-SiO₂
- Other gate insulator materials tend to have much higher interface state densities

All of the above....

- For the three types of oxide charges the CV curve is shifted by the voltage on the capacitor Q/C

$$V_{FB-oxide_charge} = \frac{-1}{C_i} \left[\frac{1}{d} \int_0^d x \rho(x) dx \right]$$

- When work function differences and oxide charges are present, the flat band voltage shift is:

$$V_{FB} = \phi_{ms} - \frac{(Q_f - Q_m - Q_{ot})}{C_i}$$

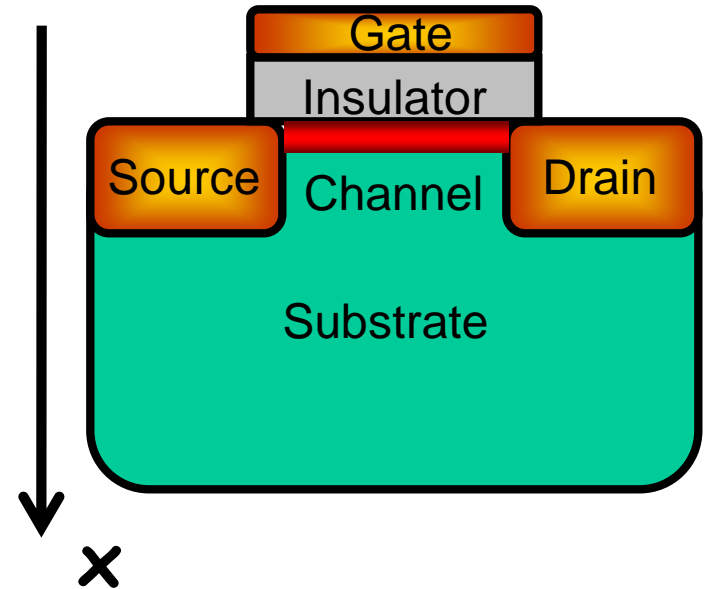
Some important equations in the inversion regime (Depth direction)

$$V_T = \phi_{ms} + 2\psi_B + \psi_{ox}$$

$$\psi_{ox} = Q_s / C_{ox}$$

$$Q_s = qN_A W_{dm}$$

$$W_{dm} = \sqrt{[2\epsilon_S(2\psi_B)/qN_A]}$$



$$V_T = \phi_{ms} + 2\psi_B + (\sqrt{[4\epsilon_S\psi_B qN_A]} - Q_f + Q_m + Q_{ot}) / C_{ox}$$

$$Q_{inv} = C_{ox}(V_G - V_T)$$